## NUMBER SYSTEM AND CODES

- The term digital refers to a process that is achieved by using discrete unit.
- In number system there are different symbols and each symbol has an absolute value and also has place value.


## RADIX OR BASE:-

The radix or base of a number system is defined as the number of different digits which can occur in each position in the number system.

## RADIX POINT :-

The generalized form of a decimal point is known as radix point. In any positional number system the radix point divides the integer and fractional part.
$\mathrm{N}_{\mathrm{r}}=$ [ Integer part .Fractional part ]
$\uparrow$
Radix point
NUMBER SYSTEM:-
In general a number in a system having base or radix ' $r$ ' can be written as

$$
\mathbf{a}_{\mathrm{n}} \quad \mathbf{a}_{\mathrm{n}-1} \quad \mathbf{a}_{\mathrm{n}-2} \ldots \ldots \ldots \ldots \ldots . . \mathbf{a}_{0} \text {. } \mathbf{a}_{-1} \mathbf{a}_{-2} . . . . . . . . . . . . . . . . . . . . \mathbf{a}_{-m}
$$

This will be interpreted as
$Y=a_{n} \times r^{n}+a_{n-1} \times r^{n-1}+a_{n-2} \times r^{n-2}+\ldots \ldots \ldots+a_{0} \times r^{0}+a_{-1} \times r^{-1}+a_{-2} \times r^{-2}+\ldots \ldots+a_{-m} \times r^{-m}$
where $\quad Y=$ value of the entire number

$$
\begin{aligned}
& a_{n}=\text { the value of the } n^{\text {th }} \text { digit } \\
& r=\text { radix }
\end{aligned}
$$

## TYPES OF NUMBER SYSTEM:-

There are four types of number systems. They are

1. Decimal number system
2. Binary number system
3. Octal number system
4. Hexadecimal number system

## DECIMAL NUMBER SYSTEM:-

- The decimal number system contain ten unique symbols $0,1,2,3,4,5,6,7,8$ and 9 .
- In decimal system 10 symbols are involved, so the base or radix is 10 .
- It is a positional weighted system.
- The value attached to the symbol depends on its location with respect to the decimal point.
$\qquad$ d - m
is given by
$\left(d_{n} \times 10^{n}\right)+\left(d_{n-1} \times 10^{n-1}\right)+\left(d_{n-2} \times 10^{n-2}\right)+\ldots+\left(d_{0} \times 10^{0}\right)+\left(d_{-1} \times 10^{-1}\right)+\left(d_{-2} \times 10^{-2}\right)+\ldots+\left(d_{-m} \times 10^{-m}\right)$


## For example:-

$$
\begin{aligned}
9256.26 & =9 \times 1000+2 \times 100+5 \times 10+6 \times 1+2 \times(1 / 10)+6 \times(1 / 100) \\
& =9 \times 10^{3}+2 \times 10^{2}+5 \times 10^{1}+6 \times 10^{0}+2 \times 10^{-1}+6 \times 10^{-2}
\end{aligned}
$$

## BINARY NUMBER SYSTEM:-

- The binary number system is a positional weighted system.
- The base or radix of this number system is 2 .
- It has two independent symbols.
- The symbols used are 0 and 1 .
- A binary digit is called a bit.
- The binary point separates the integer and fraction parts.

In general,
is given by
$\left(d_{n} \times 2^{n}\right)+\left(d_{n-1} \times 2^{n-1}\right)+\left(d_{n-2} \times 2^{n-2}\right)+\ldots+\left(d_{0} \times 2^{0}\right)+\left(d_{-1} \times 2^{-1}\right)+\left(d_{-2} \times 2^{-2}\right)+\ldots+\left(d_{-k} \times 2^{-k}\right)$

## OCTAL NUMBER SYSTEM:-

- It is also a positional weighted system.
- Its base or radix is 8 .
- It has 8 independent symbols $0,1,2,3,4,5,6$ and 7 .
- Its base $8=2^{3}$, every 3 - bit group of binary can be represented by an octal digit.


## HEXADECIMAL NUMBER SYSTEM:-

- The hexadecimal number system is a positional weighted system.
- The base or radix of this number system is 16 .
- The symbols used are $0,1,2,3,4,5,6,7,8,9, \mathrm{~A}, \mathrm{~B}, \mathrm{C}, \mathrm{D}, \mathrm{E}$ and F
- The base $16=24$, every 4 - bit group of binary can be represented by an hexadecimal digit.


## CONVERSION FROM ONE NUMBER SYSTEM TO ANOTHER :-

## 1. BINARY NUMBER SYSTEM:-

(a) Binary to decimal conversion:-

In this method, each binary digit of the number is multiplied by its positional weight and the product terms are added to obtain decimal number.

## For example:

## (i) Convert (10101)2 to decimal.

## Solution :

(Positional weight) $\quad 2^{4} 2^{3} 2^{2} 2^{1} 2^{0}$
Binary number

$$
\begin{aligned}
& 10101 \\
& =\left(1 \times 2^{4}\right)+\left(0 \times 2^{3}\right)+\left(1 \times 2^{2}\right)+\left(0 \times 2^{1}\right)+\left(1 \times 2^{0}\right) \\
& =16+0+4+0+1 \\
& =(21)_{10}
\end{aligned}
$$

## (ii) Convert (111.101) $)_{2}$ to decimal.

## Solution:

$$
\begin{aligned}
(111.101)_{2} & =\left(1 \times 2^{2}\right)+\left(1 \times 2^{1}\right)+\left(1 \times 2^{0}\right)+\left(1 \times 2^{-1}\right)+\left(0 \times 2^{-2}\right)+\left(1 \times 2^{-3}\right) \\
& =4+2+1+0.5+0+0.125 \\
& =(7.625)_{10}
\end{aligned}
$$

## (b) Binary to Octal conversion:-

For conversion binary to octal the binary numbers are divided into groups of 3 bits each, starting at the binary point and proceeding towards left and right.

| Octal | Binary | Octal | Binary |
| :---: | :---: | :---: | :---: |
| 0 | 000 | 4 | 100 |
| 1 | 001 | 5 | 101 |
| 2 | 010 | 6 | 110 |
| 3 | 011 | 7 | 111 |

For example:
(i) Convert $\mathbf{( 1 0 1 1 1 1 0 1 0 1 1 0 . 1 1 0 1 1 0 0 1 1 )})_{2}$ into octal.

## Solution :

$\begin{array}{lcccccccc}\text { Group of } 3 \text { bits are } & 101 & 111 & 010 & 110 & & 110 & 110 & 011 \\ \text { Convert each group into octal }= & 5 & 7 & 2 & 6 & . & 6 & 6 & 3\end{array}$
The result is $\mathbf{( 5 7 2 6 . 6 6 3})_{\mathbf{8}}$
(ii) Convert (10101111001.0111) $)_{2}$ into octal.

## Solution :

| Binary number | 10 | 101 | 111 | 001 |  | 011 | 1 |
| :--- | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| Group of 3 bits are | $=010$ | 101 | 111 | 001 | . | 011 | 100 |
| Convert each group into octal $=$ | 2 | 5 | 7 | 1 | . | 3 | 4 |

The result is (2571.34)8

## (c) Binary to Hexadecimal conversion:-

For conversion binary to hexadecimal number the binary numbers starting from the binary point, groups are made of 4 bits each, on either side of the binary point.

| Hexadecimal | Binary | Hexadecimal | Binary |
| :---: | :---: | :---: | :---: |
| 0 | 0000 | 8 | 1000 |
| 1 | 0001 | 9 | 1001 |
| 2 | 0010 | A | 1010 |
| 3 | 0011 | B | 1011 |
| 4 | 0100 | C | 1100 |
| 5 | 0101 | D | 1101 |
| 6 | 0110 | E | 1110 |
| 7 | 0111 | F | 1111 |

## For example:

(i) Convert (1011011011) $)_{2}$ into hexadecimal.

## Solution:

| Given Binary number | 10 | 1101 | 1011 |
| :--- | ---: | :---: | :---: |
| Group of 4 bits are | 0010 | 1101 | 1011 |
| Convert each group into hex | $=$ | 2 | D |

The result is (2DB) ${ }_{16}$
(ii) Convert ( 01011111011.011111$)_{2}$ into hexadecimal.

## Solution:

$\begin{array}{lrlllll}\text { Given Binary number } & 010 & 1111 & 1011 & \text {. } & 0111 & 11 \\ \text { Group of } 3 \text { bits are } & =0010 & 1111 & 1011 & . & 0111 & 1100\end{array}$
Convert each group into octal $=\begin{array}{llllll} & 2 & F & B & \end{array}$
The result is (2FB.7C) ${ }_{16}$

## 2. DECIMAL NUMBER SYSTEM:-

## (a) Decimal to binary conversion:-

In the conversion the integer number are converted to the desired base using successive division by the base or radix.
For example:
(i) Convert (52) ${ }_{10}$ into binary.

## Solution:

Divide the given decimal number successively by 2 read the integer part remainder upwards to get equivalent binary number. Multiply the fraction part by 2. Keep the integer in the product as it is and multiply the new fraction in the product by 2 . The process is continued and the integer are read in the products from top to bottom.

$$
\begin{array}{ll}
2 \underline{I 52} & \\
2 \underline{126} & -0 \\
2 \underline{113} & -0 \\
2 \underline{16} & -1 \\
2 \underline{13} & -0 \\
2 \underline{11} & -1 \\
0 & -1
\end{array}
$$

(ii) Convert (105.15) ${ }_{10}$ into binary.

## Solution:

| Integer <br> part | Fraction part |
| :--- | :--- |
| $2 \underline{\underline{1} 105}$ |  |
| $2 \underline{\underline{152}}-1$ | $0.15 \times 2=0.30$ |
| $2 \underline{126}$ | -0 |
| $2 \underline{\underline{113}}$ | -0 |
| $2 \underline{16}$ | -1 |

Result of $(105.15)_{10}$ is $(\mathbf{1 1 0 1 0 0 1 . 0 0 1 0 0 1})_{\mathbf{2}}$

## (b) Decimal to octal conversion:-

To convert the given decimal integer number to octal, successively divide the given number by 8 till the quotient is 0 . To convert the given decimal fractions to octal successively multiply the decimal fraction and the subsequent decimal fractions by 8 till the product is 0 or till the required accuracy is obtained.

For example:
(i) Convert (378.93) ${ }_{10}$ into octal.

Solution:
$8 \longdiv { 3 7 8 }$
$0.93 \times 8=7.44$
$8 \underline{\mid 47}-2$
$0.44 \times 8=3.52$
$\begin{array}{rr}8 \underline{\mid 5} & -7 \\ 0 & -5\end{array}$
$0.52 \times 8=4.16$
$0.16 \times 8=1.28$

Result of (378.93 $)_{10}$ is $\left.\mathbf{( 5 7 2 . 7 3 4 1}\right)_{8}$

## (c) Decimal to hexadecimal conversion:-

The decimal to hexadecimal conversion is same as octal.

## For example:

(i) Convert (2598.675) $)_{10}$ into hexadecimal.

## Solution:

| Remainder <br> Decimal <br> Hex |  |  |  | Hex |
| :--- | :--- | :--- | :--- | :---: |
| $16 \underline{\mathrm{I} 2598}$ |  |  | $0.675 \times 16=10.8$ | A |
| $16 \underline{\underline{1162}}$ | -6 | 6 | $0.800 \times 16=12.8$ | C |
| $16 \underline{110}$ | -2 | 2 | $0.800 \times 16=12.8$ | C |
| 0 | -10 | A | $0.800 \times 16=12.8$ | C |

Result of $(2598.675)_{10}$ is $(\mathbf{A 2 6 . A C C C})_{16}$

## 3. OCTAL NUMBER SYSTEM:-

(a) Octal to binary conversion:-
b. To convert a given a octal number to binary, replace each octal digit by its 3-bit binary equivalent.

## For example:

Convert (367.52)8 into binary.
Solution:
$\begin{array}{lcccccc}\text { Given Octal number is } & 3 & 6 & 7 & \text { • } & 5 & 2 \\ \text { Convert }\end{array}$ to binary

Result of $(367.52)_{8}$ is $\left.\mathbf{( \mathbf { 0 1 1 1 1 0 1 1 1 }} \mathbf{1 0 1 0 1 0}\right)_{2}$

## (b) Octal to decimal conversion:-

For conversion octal to decimal number, multiply each digit in the octal number by the weight of its position and add all the product terms

## For example: -

Convert (4057.06) 8 to decimal

## Solution:

$$
\begin{aligned}
(4057.06)_{8} & =4 \times 8^{3}+0 \times 8^{2}+5 \times 8^{1}+7 \times 8^{0}+0 \times 8^{-1}+6 \times 8^{-2} \\
& =2048+0+40+7+0+0.0937 \\
& =(2095.0937)_{10}
\end{aligned}
$$

Result is $\mathbf{( 2 0 9 5 . 0 9 3 7})_{10}$
(c) Octal to hexadecimal conversion:-

For conversion of octal to Hexadecimal, first convert the given octal number to binary and then binary number to hexadecimal.

## For example :- <br> Convert (756.603)8 to hexadecimal.

Solution :-
Given octal no.
Convert each octal digit to binary
Group of 4bits are
Convert 4 bits group to hex.

|  | 7 | 5 | 6 | $\cdot$ | 6 | 0 | 3 |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $=$ | 111 | 101 | 110 | $\cdot$ | 110 | 000 | 011 |
| $=$ | 0001 | 1110 | 1110 | $\cdot$ | 1100 | 0001 | 1000 |
| $=$ | 1 | E | E | $\cdot$ | C | 1 | 8 |

Result is (1EE.C18) ${ }_{\mathbf{1 6}}$

## (4) HEXADECIMAL NUMBER SYSTEM :-

(a) Hexadecimal to binary conversion:-

For conversion of hexadecimal to binary, replace hexadecimal digit by its 4 bit binary group.
For example:
Convert (3A9E.B0D) ${ }_{16}$ into binary.

## Solution:

Given Hexadecimal number is

$$
\begin{array}{cccccccc}
3 & \text { A } & 9 & \text { E } & \text { B } & 0 & \text { D } \\
=0011 & 1010 & 1001 & 1110 & . & 1011 & 0000 & 1101
\end{array}
$$

Convert each hexadecimal to 4 bit binary

Result of (3A9E.B0D) $)_{8}$ is $\left.\mathbf{( 0 0 1 1 1 0 1 0 1 0 0 1 1 1 1 0 . 1 0 1 1 0 0 0 0 1 1 0 1}\right)_{2}$

## (b) Hexadecimal to decimal conversion:-

For conversion of hexadecimal to decimal, multiply each digit in the hexadecimal number by its position weight and add all those product terms.

## For example: -

Convert (A0F9.0EB) $\mathbf{1 6}_{6}$ to decimal

## Solution:

$$
\begin{aligned}
(\mathrm{A} 0 \mathrm{~F} 9.0 \mathrm{~EB})_{16} & =\left(10 \times 16^{3}\right)+\left(0 \times 16^{2}\right)+\left(15 \times 16^{1}\right)+\left(9 \times 16^{0}\right)+\left(0 \times 16^{-1}\right)+\left(14 \times 16^{-2}\right)+\left(11 \times 16^{-3}\right) \\
& =40960+0+240+9+0+0.0546+0.0026 \\
& =(41209.0572)_{10}
\end{aligned}
$$

Result is $\mathbf{( 4 1 2 0 9 . 0 5 7 2})_{\mathbf{1 0}}$
(c) Hexadecimal to Octal conversion:-

For conversion of hexadecimal to octal, first convert the given hexadecimal number to binary and then binary number to octal.

For example :-
Convert (B9F.AE) ${ }_{16}$ to octal.

## Solution :-

Given hexadecimal no.is
Convert each hex. digit to binary $\quad=\quad \begin{array}{llllll}1011 & 1001 & 1111 & . & 1010 & 1110\end{array}$
Group of 3 bits are $\quad=\quad \begin{array}{lllllllllllllll}101 & 110 & 011 & 111 & & 101 & 011 & 100\end{array}$
$\begin{array}{llllllllllll}\text { Convert } 3 \text { bits group to octal. } & = & 5 & 6 & 3 & 7 & & 5 & 3 & 4\end{array}$
Result is $\mathbf{( 5 6 3 7 . 5 3 4 )} \mathbf{)}_{\mathbf{8}}$

## BINARY ARITHEMATIC OPERATION :-

## 1. BINARY ADDITION:-

The binary addition rules are as follows
$0+0=0 ; 0+1=1 ; 1+0=1 ; 1+1=10$, i.e 0 with a carry of

## 1 For example :-

Add (100101)2 ${ }_{2}$ and (1101111)2.
Solution :-

$$
\begin{array}{r}
100101 \\
+\quad 1101111 \\
\hline 10010100 \\
\hline
\end{array}
$$

Result is $(\mathbf{1 0 0 1 0 1 0 0})_{2}$

## 2. BINARY SUBTRACTION:-

The binary subtraction rules are as follows
$0-0=0 ; 1-1=0 ; 1-0=1 ; 0-1=1$, with a borrow of 1

For example :-
Substract (111.111) $\mathbf{2}_{2}$ from (1010.01)2.
Solution :-

$$
\begin{array}{r}
1010.010 \\
-\quad 111.111 \\
\hline 0010.011 \\
\hline
\end{array}
$$

Result is $\mathbf{( 0 0 1 0 . 0 1 1 )} \mathbf{2}_{2}$

## 3. BINARY MULTIPLICATION:-

The binary multiplication rules are as
follows $0 \times 0=0 ; 1 \times 1=1 ; 1 \times 0=0 ; 0 \times$
$1=0$
For example :-
Multiply (1101)2 by (110)2. Solution :-

$$
\begin{array}{r}
1101 \\
\times \quad 110 \\
\hline \\
0000 \\
\\
+\quad 1101 \\
\hline 1001 \\
\hline
\end{array}
$$

Result is $\mathbf{( 1 0 0 1 1 1 0 )})_{2}$

## 4. BINARY DIVISION:-

The binary division is very simple and similar to decimal number system. The division by ' 0 ' is meaningless. So we have only 2 rules
$0 \div 1=0$
$1 \div 1=1$
For example :-
Divide (10110) $)_{2}$ by (110) ${ }_{2}$.
Solution :-
110) 101101 ( 111.1

- 110

1010
$\qquad$
1001
110
110
$\qquad$
Result is $\mathbf{( 1 1 1 . 1 )} \mathbf{I}_{\mathbf{2}}$

## 1's COMPLEMENT REPRESENTATION :-

The 1 's complement of a binary number is obtained by changing each 0 to 1 and each 1 to 0 .

## For example :-

Find (1100) $)_{2}$ 1's complement.

## Solution :-

| Given | 1 | 1 | 0 | 0 |
| :---: | :---: | :---: | :---: | :---: |
| 1 's complement is | 0 | 0 | 1 | 1 |

Result is (0011) $\mathbf{2}_{2}$

## 2's COMPLEMENT REPRESENTATION:-

The 2 's complement of a binary number is a binary number which is obtained by adding 1 to the 1 's complement of a number i.e.

$$
2 \text { 's complement }=1 \text { 's complement }+1
$$

## For example :-

Find (1010) $)_{2}$ 2's complement.
Solution :-

| Given | 1 | 0 | 1 | 0 |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 1's complement is |  | 0 | 1 | 0 | 1 |
|  | + |  |  |  | 1 |

2's complement


Result is (0110) $\mathbf{2}_{2}$

## SIGNED NUMBER :-

In sign - magnitude form, additional bit called the sign bit is placed in front of the number. If the sign bit is 0 , the number is positive. If it is a 1 , the number is negative.

## For example:-

$$
\begin{array}{ccccccccc}
0 & 1 & 0 & 1 & 0 & 0 & 1 & =+41 \\
\uparrow \\
& & & & & & & \\
\text { Sign bit } & & & & & &
\end{array}
$$

```
    1
    \uparrow
Sign bit
```


## SUBSTRACTION USING COMPLEMENT METHOD :-

## 1's COMPLEMENT:-

In 1's complement subtraction, add the 1's complement of subtrahend to the minuend. If there is a carry out, then the carry is added to the LSB. This is called end around carry. If the MSB is 0 , the result is positive. If the MSB is 1 , the result is negative and is in its 1 's complement form. Then take its 1 's complement to get the magnitude in binary.

## For example:-

Subtract (10000) $)_{2}$ from (11010) $)_{2}$ using 1 's complement.

## Solution:-

| 11010 |  |  | 11010 |  | $=26$ |
| :---: | :---: | :---: | :---: | :---: | :---: |
| -10000 | => |  | +01111 | (1's complement) | $=-\underline{16}$ |
|  | Carry |  | 101001 |  | + 10 |
|  |  |  | 1 |  |  |
|  |  |  | 01010 | $=+10$ |  |

Result is $\mathbf{+ 1 0}$

## 2's COMPLEMENT:-

In 2's complement subtraction, add the 2 's complement of subtrahend to the minuend. If there is a carry out, ignore it. If the MSB is 0 , the result is positive. If the MSB is 1 , the result is negative and is in its 2 's complement form. Then take its 2's complement to get the magnitude in binary.

## For example:-

## Subtract (1010100) $)_{2}$ from (1010100) $)_{2}$ using 2's complement.

Solution:-
1010100

1010100 $\quad \Rightarrow \quad$| 1010100 | $=84$ |
| :---: | :---: |
| $+0101100(2$ s complement $)$ | $=-\underline{84}$ |

Hence MSB is 0 . The answer is positive. So it is $+0000000=\mathbf{0}$

## DIGITAL CODES:-

In practice the digital electronics requires to handle data which may be numeric, alphabets and special characters. This requires the conversion of the incoming data into binary format before it can be processed. There is various possible ways of doing this and this process is called encoding. To achieve the reverse of it, we use decoders.

## WEIGHTED AND NON-WEIGHTED CODES:-

There are two types of binary codes

1) Weighted binary codes
2) Non- weighted binary codes

In weighted codes, for each position ( or bit) ,there is specific weight attached.
For example, in binary number, each bit is assigned particular weight 2 n where ' n ' is the bit number for $\mathrm{n}=$ $0,1,2,3,4$ the weights are $1,2,4,8,16$ respectively.
Example :- BCD
Non-weighted codes are codes which are not assigned with any weight to each digit position, i.e., each digit position within the number is not assigned fixed value.
Example:- Excess - 3 (XS -3) code and Gray codes

## BINARY CODED DECIMAL (BCD):-

BCD is a weighted code. In weighted codes, each successive digit from right to left represents weights equal to some specified value and to get the equivalent decimal number add the products of the weights by the corresponding binary digit. 8421 is the most common because 8421 BCD is the most natural amongst the other possible codes.

## For example:-

$(567)_{10}$ is encoded in various 4 bit codes.

## Solution:-

| Decimal | $\rightarrow$ | 5 | 6 | 7 |
| :--- | :--- | :---: | :---: | :---: |
| 8421 code | $\rightarrow$ | 0101 | 0110 | 0111 |
| 6311 code | $\rightarrow$ | 0111 | 1000 | 1001 |
| 5421 code | $\rightarrow$ | 1000 | 0100 | 1010 |

## BCD ADDITION:-

Addition of BCD (8421) is performed by adding two digits of binary, starting from least significant digit. In case if the result is an illegal code (greater than 9) or if there is a carry out of one then add $0110(6)$ and add the resulting carry to the next most significant.

## For example:-

Add 679.6 from 536.8 using BCD addition.

## Solution:-

| 679.6 | 011001111001 . 0110 | ( 679.6 in BCD) |
| :---: | :---: | :---: |
| +536.8 | $=>+\underline{0101} 00011 \quad 0110$ • 1000 | (536.8 in BCD) |
| 1216.4 | 101110101111 . 1110 | ( All are illegal codes) |
|  | $+0110+0110+0110 .+0110$ | ( Add 0110 to each) |
|  | 0001001000010110.0100 |  |
|  | 216 | $($ corrected sum $=1216.4)$ |

Result is $\mathbf{1 2 1 6 . 4}$

## BCD SUBTRACTION:-

The BCD subtraction is performed by subtracting the digits of each 4 - bit group of the subtrahend from corresponding 4 - bit group of the minuend in the binary starting from the LSD. If there is no borrow from the next higher group[ then no correction is required. If there is a borrow from the next group, then $6_{10}(0110)$ is subtracted from the difference term of this group.
For example:-
Subtract 147.8 from 206.7 using 8421 BCD code.

## Solution:-



Result is (58.9) ${ }_{\mathbf{1 0}}$

## EXCESS THREE(XS-3) CODE:-

The Excess-3 code, also called XS-3, is a non- weighted BCD code. This derives it name from the fact that each binary code word is the corresponding 8421 code word plus $0011(3)$. It is a sequential code. It is a self complementing code.

## XS-3 ADDITION:-

In XS-3 addition, add the XS-3 numbers by adding the 4 bit groups in each column starting from the LSD. If there is no carry out from the addition of any of the 4 bit groups, subtract 0011 from the sum term of those groups. If there is a carry out, add 0011 to the sum term of those groups

## For example:-

Add 37 and 28 using XS-3 code.

## Solution:-

| 37 |  | 01101010 | ( 37 in XS-3) |
| :---: | :---: | :---: | :---: |
| $+\underline{28}$ | => | + 01011011 | ( 28 in XS-3) |
| 65 |  | $\begin{aligned} & 101111010 \\ & +\quad 1 \\ & \hline \end{aligned}$ | ( Carry is generated) <br> ( Propagate carry) |
|  |  | 11000101 | ( Add 0110 to correct 0101 and |
|  |  | $-\underline{0011+0011}$ | subtract 0011 to correct 1100) |
|  |  | 10011000 | ( Corrected sum in XS-3 $=65_{10}$ ) |

## XS-3 SUBTRACTION:-

To subtract in XS-3 number by subtracting each 4-bit group of the subtrahend from the corresponding 4-bit group of the minuend starting from the LSD. If there is no borrow from the next 4-bit group. add 0011 to the difference term of such groups. If there is a borrow, subtract 0011 from the difference term.

## For example :-

. Subtract 175 from 267 using XS-3 code.

## Solution :-`

$$
\begin{aligned}
& \underline{-175} \quad \Rightarrow \quad-\underline{0100 \quad 1010 \quad 1000} \text { ( } 175 \text { in XS-3) } \\
& \begin{array}{rrrrr} 
& \begin{array}{lll}
0101 & 1010 & 1010
\end{array} & (267 \text { in XS-3) } \\
\Rightarrow & -\underline{0100} & 1010 & 1000 & (175 \text { in XS-3) }
\end{array} \\
& 000011110010 \text { (Correct } 0010 \text { and } 0000 \text { by adding } 0011 \text { and } \\
& +0011-0011+0011 \text { correct } 1111 \text { by subtracting 0011) } \\
& 001111000101 \text { (Corrected difference in XS-3 = 92 }{ }_{10} \text { ) }
\end{aligned}
$$

## ASCII CODE:-

The American Standard Code for Information Interchange (ASCII) pronounced as 'ASKEE' is widely used alphanumeric code. This is basically a 7 bit code. The number of different bit patterns that can be created with 7 bits is $27=128$, the ASCII can be used to encode both the uppercase and lowercase characters of the alphabet ( 52 symbols) and some special symbols in addition to the 10 decimal digits. It is used extensively for printers and terminals that interface with small computer systems. The table shown below shows the ASCII groups.

The ASCII code

| LSBs | MSBs |  |  |  |  |  |  |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
|  | 000 | 001 | 010 | 011 | 100 | 101 | 110 | 11 |
| 0000 | NUL | DEL | Space | 0 | @ | P | P |  |
| 0001 | SOH | DC1 | $!$ | 1 | A | Q | a | q |
| 0010 | STX | DC2 | $"$ | 2 | B | R | b | r |
| 0011 | ETX | DC3 | $\#$ | 3 | C | S | c | s |


| 0100 | EOT | DC4 | $\$$ | 4 | D | T | d | t |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 0101 | ENQ | NAK | $\%$ | 5 | E | U | e | u |
| 0110 | ACK | SYN | \& | 6 | F | V | f | v |
| 0111 | BEL | ETB | $\cdot$ | 7 | G | W | g | w |
| 1000 | BS | CAN | $($ | 8 | H | X | h | x |
| 1001 | HT | EM | $)$ | 9 | I | Y | i | y |
| 1010 | LF | SUB | $*$ | $:$ | J | Z | j | z |
| 1011 | VT | ESC | + | $;$ | K | [ | k |  |
| 1100 | FF | FS | , | < | L | l | l | l |
| 1101 | CR | GS | - | $=$ | M | l | m | \} |
| 1110 | SO | RS | $\cdot$ | $>$ | N | $\wedge$ | n | $\sim$ |
| 1111 | SI | US | l | $?$ | O | - | o | DLE |

## EBCDIC CODE:-

The Extended Binary Coded Decimal Interchange Code (EBCDIC) pronounced as 'eb - si- dik' is an 8 bit alphanumeric code. Since $28=256$ bit patterns can be formed with 8 bits. It is used by most large computers to communicate in alphanumeric data. The table shown below shows the EBCDIC code.

The EBCDIC code

| LSD | MSD(Hex) |  |  |  |  |  |  |  |  |  |  |  |  |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | 0 | 1 | 2 | 3 | 4 | 5 | 6 | 7 | 8 | 9 | A | B | C | D | E | F |
| 0 | NUL | DLE | DS |  | SP | \& |  |  |  |  |  |  | [ | ] | I | 0 |
| 1 | SOH | DC1 | SOS |  |  |  | 1 |  | a | j | $\sim$ |  | A | J |  | 1 |
| 2 | STX | DC2 | FS | SYN |  |  |  |  | b | k | S |  | B | K | S | 2 |
| 3 | ETX | DC3 |  |  |  |  |  |  | c | 1 | t |  | C | L | T | 3 |
| 4 | PF | RES | BYP | PN |  |  |  |  | d | m | u |  | D | M | U | 4 |
| 5 | HT | NL | LF | RS |  |  |  |  | e | n | v |  | E | N | V | 5 |
| 6 | LC | BS | EOB | YC |  |  |  |  | f | o | w |  | F | O | W | 6 |
| 7 | DEL | IL | PRE | EOT |  |  |  |  | g | p | X |  | G | P | X | 7 |
| 8 |  | CAN |  |  |  |  |  |  | h | q | y |  | H | Q | Y | 8 |
| 9 |  | EM |  |  |  |  |  |  | i | r | Z |  | I | R | Z | 9 |
| A | SMM | CC | SM |  | Ø | $!$ | I | : |  |  |  |  |  |  |  |  |
| B | VT |  |  |  | - | \$ | , | \# |  |  |  |  |  |  |  |  |
| C | FF | IFS |  | DC4 | $<$ | * | \% | @ |  |  |  |  |  |  |  |  |
| D | CR | IGS | ENQ | NAK | ( | ) | - | ' |  |  |  |  |  |  |  |  |
| E | SO | IRS | ACK |  | + | ; | > | $=$ |  |  |  |  |  |  |  |  |
| F | SI | IUS | BEL | SUB | I | ' | ? | ' |  |  |  |  |  |  |  |  |

## GRAY CODE:-

The gray code is a non-weighted code. It is not a BCD code. It is cyclic code because successive words in this differ in one bit position only i.e it is a unit distance code.

Gray code is used in instrumentation and data acquisition systems where linear or angular displacement is measured. They are also used in shaft encoders, I/O devices, A/D converters and other peripheral equipment.

## BINARY- TO - GRAY CONVERSION:-

If an n-bit binary number is represented by $B_{n} B_{n-1}---B_{1}$ and its gray code equivalent by $G_{n} G_{n-1}--$ G1, where $B_{n}$ and $G_{n}$ are the MSBs, then gray code bits are obtained from the binary code as follows

$$
\begin{aligned}
& \mathrm{G}_{\mathrm{n}}=\mathrm{B}_{\mathrm{n}} \\
& \mathrm{G}_{\mathrm{n}-1}=\mathrm{B}_{\mathrm{n}}^{\oplus} \mathrm{B}_{\mathrm{n}-1} \\
& \cdot \\
& \cdot \\
& \cdot \\
& \cdot \\
& \cdot \\
& \mathrm{G}_{1}=\mathrm{B}_{2} \oplus_{\mathrm{B}_{1}}
\end{aligned}
$$

Where the symbol ${ }^{\oplus}$ stands for Exclusive OR (X-OR)

## For example :-

Convert the binary 1001 to the Gray code.

## Solution :-`



The gray code is $\mathbf{1 1 0 1}$

## GRAY- TO - BINARY CONVERSION:-

If an n-bit gray number is represented by $G_{n} G_{n-1}------G_{1}$ and its binary equivalent by $B_{n} B_{n-1}-----\quad B 1$, then binary bits are obtained from Gray bits as follows

$$
\begin{aligned}
&: \mathrm{B}_{\mathrm{n}}= \mathrm{G}_{\mathrm{n}} \\
& \mathrm{~B}_{\mathrm{n}-1}=\mathrm{B}_{\mathrm{P} P} \mathrm{G}_{\mathrm{n}-1} \\
& \cdot \\
& \cdot \\
& \cdot \\
& \cdot \\
& \mathrm{~B}_{1}=\mathrm{B}_{2} \oplus{ }_{\mathrm{G}}^{1}
\end{aligned}
$$

## For example :-

Convert the Gray code 1101 to the binary.

## Solution :-



The binary code is $\mathbf{1 0 0 1}$

## LOGIC GATES

## LOGIC GATES:-

- Logic gates are the fundamental building blocks of digital systems.
- There are 3 basic types of gates AND, OR and NOT.
- Logic gates are electronic circuits because they are made up of a number of electronic devices and components.
- Inputs and outputs of logic gates can occur only in 2 levels. These two levels are termed HIGH and LOW, or TRUE and FALSE, or ON and OFF or simply 1 and 0.
- The table which lists all the possible combinations of input variables and the corresponding outputs is called a truth table.


## LEVEL LOGIC:-

A logic in which the voltage levels represents logic 1 and logic 0 . Level logic may be positive or negative logic. Positive Logic:-
A positive logic system is the one in which the higher of the two voltage levels represents the logic 1 and the lower of the two voltages level represents the logic 0 .

## Negative Logic:-

A negative logic system is the one in which the lower of the two voltage levels represents the logic 1 and the higher of the two voltages level represents the logic 0 .

## DIFFERENT TYPES OF LOGIC GATES:- NOT

## GATE (INVERTER):-

- A NOT gate, also called and inverter, has only one input and one output.
- It is a device whose output is always the complement of its input.
- The output of a NOT gate is the logic 1 state when its input is in logic 0 state and the logic 0 state when its inputs is in logic 1 state.

IC No. :- 7404


## Truth table

| INPUT <br> $\mathbf{A}$ | OUTPUT <br> $\mathbf{A}$ |
| :---: | :---: |
| 0 | $\mathbf{1}$ |
| 1 | 0 |

## Timing Diagram

A


A

$\begin{array}{llll}0 & 1 & 1 & 0\end{array}$

- An AND gate has two or more inputs but only one output.
- The output is logic 1 state only when each one of its inputs is at logic 1 state.
- The output is logic 0 state even if one of its inputs is at logic 0 state.


## IC No.:- 7408

Logic Svmbol


Timing Diagram
0
A


Truth Table

|  |  | OUTPUT |
| :---: | :---: | :---: |
| A | B | Q=A. B |
| 0 | 0 | 0 |
| 0 | 1 | 0 |
| 1 | 0 | 0 |
| 1 | 1 | 1 |

## OR GATE:-

- An OR gate may have two or more inputs but only one output.
- The output is logic 1 state, even if one of its input is in logic 1 state.
- The output is logic 0 state, only when each one of its inputs is in logic state.


## IC No.:- 7432

Logic Symbol
Truth Table


Timing Diagram

| INPUT |  | OUTPUT |
| :---: | :---: | :---: |
| $\mathbf{A}$ | $\mathbf{B}$ | Q=A + B |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 1 |

$0 \quad 0$
$1 \quad 1$

A


B



- NAND gate is a combination of an AND gate and a NOT gate.
- The output is logic 0 when each of the input is logic 1 and for any other combination of inputs, the output is logic 1 .


## IC No.:- 7400 two input NAND gate <br> 7410 three input NAND <br> gate 7420 four input NAND <br> gate 7430 eight input NAND <br> gate

## Logic Symbol



## Timing Diagram

$\begin{array}{llll}\mathbf{0} & \mathbf{0} & \mathbf{1} & \mathbf{1}\end{array}$

A


Q

NOR GATE:-

- NOR gate is a combination of an OR gate and a NOT gate.
- The output is logic 1 , only when each one of its input is logic 0 and for any other combination of inputs, the output is a logic 0 level.

IC No.:- 7402 two input NOR gate
7427 three input NOR gate
7425 four input NOR gate

Logic Symbol
Truth Table


| INPUT |  | OUTPUT |
| :---: | :---: | :---: |
| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{Q = A}+\mathbf{B}$ |
| 0 | 0 | 1 |
| 0 | 1 | 0 |
| 1 | 0 | 0 |
| 1 | 1 | 0 |

## Timing Diagram

$\begin{array}{llll}\mathbf{0} & \mathbf{0} & \mathbf{1} & \mathbf{1}\end{array}$
A

B

Q


## EXCLUSIVE - OR (X-OR) GATE:-

- An X-OR gate is a two input, one output logic circuit.
- The output is logic 1 when one and only one of its two inputs is logic 1 . When both the inputs is logic 0 or when both the inputs is logic 1 , the output is logic 0 .


## IC No.:- 7486

## Logic Symbol



INPUTS are A and B

$$
\begin{aligned}
\text { OUTPUT is } \mathbf{Q} & =\mathrm{A} \bigoplus \mathrm{~B} \\
& =\stackrel{-}{\mathrm{B}}+\overline{\mathrm{A} \mathrm{~B}}
\end{aligned}
$$

| INPUT |  | OUTPUT |
| :---: | :---: | :---: |
| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{Q}=\mathbf{A} \oplus \mathbf{B}$ |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 0 |

## Timing Diagram

$\begin{array}{llll}\mathbf{0} & \mathbf{0} & 1 & 1\end{array}$

A


B


Q


## EXCLUSIVE - NOR (X-NOR) GATE:-

- An X-NOR gate is the combination of an X-OR gate and a NOT gate.
- An X-NOR gate is a two input, one output logic circuit.
- The output is logic 1 only when both the inputs are logic 0 or when both the inputs is 1 .
- The output is logic 0 when one of the inputs is logic 0 and other is 1 .

IC No.:- 74266

## Logic Symbol



| INPUT |  | OUTPUT |
| :--- | :--- | :--- |
| $A$ | $B$ | OUT =A XNOR $B$ |
| 0 | 0 | 1 |
| 0 | 1 | 0 |
| 1 | 0 | 0 |
| 1 | 1 | 1 |

$O U T=A \quad B+A \quad B$
$=A$ XNOR B
Timing Diagram

A

$\begin{array}{llll}0 & 1 & 0 & 1\end{array}$

B


OUT


## UNIVERSAL GATES:-

There are 3 basic gates AND, OR and NOT, there are two universal gates NAND and NOR, each of which can realize logic circuits single handedly. The NAND and NOR gates are called universal building blocks. Both NAND and NOR gates can perform all logic functions i.e. AND, OR, NOT, EXOR and EXNOR. NAND GATE:-
a) Inverter from NAND gate

Input $=\mathbf{A}$
Output $\mathbf{Q}=\overline{\mathbf{A}}$
b) AND gate from NAND gate

Input $s$ are $\mathbf{A}$ and $\mathbf{B}$
Output $\mathbf{Q}=\mathbf{A} . \mathbf{B}$

c) OR gate from NAND gate

Inputs are $\mathbf{A}$ and $\mathbf{B}$
Output $\mathbf{Q}=\mathbf{A}+\mathbf{B}$

d) NOR gate from NAND gate

Inputs are $\underline{\mathbf{A} \text { and } \mathbf{B}}$
Output $\mathbf{Q}=\mathbf{A}+\mathbf{B}$

e) EX-OR gate from NAND gate

Inputs are $\mathbf{A}$ and $\mathbf{B}$
Output $\mathbf{Q}=\mathbf{A} \mathbf{B}+$
AB

f) EX-NOR gate From NAND gate

Inputs are $\mathbf{A}$ and $\mathbf{B}$
Output $\mathbf{Q}=\mathbf{A} \mathbf{B}+\overline{\mathbf{A}} \mathbf{B}$


## NOR GATE:-

a) Inverter from NOR
gate Input $=\underline{\mathbf{A}}$
Output $\mathbf{Q}=\mathbf{A}$
b) AND gate from NOR gate Input s are $\mathbf{A}$ and $\mathbf{B}$

Output $\mathbf{Q}=\mathbf{A} . \mathbf{B}$
c) OR gate from NOR gate

Inputs are $\mathbf{A}$ and $\mathbf{B}$
Output $\mathbf{Q}=\mathbf{A}+\mathbf{B}$

d) NAND gate from NOR gate

Inputs are $\boldsymbol{A}$ and $\mathbf{B}$
Output $\mathbf{Q}=\overline{\mathbf{A} . \mathbf{B}}$

e) EX-OR gate from NOR gate

Inputs are $\mathbf{A}$ and $\mathbf{B}$
Output $\mathbf{Q}=\mathbf{A B}+$
AB


## f) EX-NOR gate From NOR gate

Inputs are $\mathbf{A}$ and $\mathbf{B}$
Output $\mathbf{Q}=\mathbf{A} \mathbf{B}+\mathbf{A} \mathbf{B}$


## THRESHOLD LOGIC:-

## INTRODUCTION:-

- The threshold element, also called the threshold gate (T-gate) is a much more powerful device than any of the conventional logic gates such as NAND, NOR and others.
- Complex, large Boolean functions can be realized using much fewer threshold gates.
- Frequently a single threshold gate can realize a very complex function which otherwise might require a
large number of conventional gates.
- T-gate offers incomparably economical realization; it has not found extensive use with the digital system designers mainly because of the following limitations.

1. It is very sensitive to parameter variations.
2. It is difficult to fabricate it in IC form.
3. The speed of switching of threshold elements in much lower than that of conventional gates.

## THE THRESHOLD ELEMENTS:-

- A threshold element or gate has ' $n$ ' binary inputs $x_{1}, x_{2}, \ldots ., x_{n}$; and a single binary output $F$. But in addition to those, it has two more parameters.
- Its parameters are a threshold T and weights $\mathrm{w}_{1}, \mathrm{w}_{2}, \ldots, \mathrm{w}_{\mathrm{n}}$. The weights $\mathrm{w}_{1}, \mathrm{w}_{2}, \ldots, \mathrm{w}_{\mathrm{n}}$ are associated with the input variables $\mathrm{x}_{1}, \mathrm{x}_{2}, \ldots, \mathrm{x}_{\mathrm{n}}$.
- The value of the threshold (T) and weights may be real, positive or negative number.
- The symbol of the threshold element is shown in fig.(a).
- It is represented by a circle partitioned into two parts, one part represents the weights and other represents T.
- It is defined as
$F\left(x_{1}, x_{2}, \ldots \ldots, x_{n}\right)=1$ if and only if $\sum_{i=1}^{n} w_{i} x_{i} \geq T$
otherwise
$\mathrm{F}\left(\mathrm{x}_{1}, \mathrm{x}_{2}, \ldots \ldots, \mathrm{x}_{\mathrm{n}}\right)=0$
n
- The sum and product operation are normal arithmetic operations and the sum $\sum \mathrm{w}_{\mathrm{i}} \mathrm{X}_{\mathrm{i}} \geq \mathrm{T}$ $\mathrm{i}=1$
is called the weighted sum of the element or gate.



## Example:-

Obtain the minimal Boolean expression from the threshold gate shown in figure.


## Solution:-

The threshold gate with three inputs $\mathrm{x}_{1}, \mathrm{x}_{2}, \mathrm{x}_{3}$ with weights $-2\left(\mathrm{w}_{1}\right), 4\left(\mathrm{w}_{2}\right)$ and $2\left(\mathrm{w}_{3}\right)$ respectively. The value of threshold is $2(\mathrm{~T})$. The table shown is the weighted sums and outputs for all input combinations. For this threshold gate, the weighted sum is

$$
\begin{aligned}
\mathrm{w} & =\mathrm{w}_{1} \mathrm{x}_{1}+\mathrm{w}_{2} \mathrm{x}_{2}+\mathrm{w}_{3} \mathrm{x}_{3} \\
& =(-2) \mathrm{x}_{1}+(4) \mathrm{x}_{2}+(2) \mathrm{x}_{3} \\
& =-2 \mathrm{x}_{1}+4 \mathrm{x}_{2}+2 \mathrm{x}_{3}
\end{aligned}
$$

The output F is logic 1 for $\mathrm{w} \geq 2$ and it is logic 0 for $\mathrm{w}<2$

| Input Variables |  |  | Weighted Sum | Output |
| :--- | :--- | :--- | :--- | :--- |
| $\mathbf{x}_{\mathbf{1}}$ | $\mathbf{x}_{\mathbf{2}}$ | $\mathbf{x}_{\mathbf{3}}$ | $\mathbf{w}=\mathbf{- 2} \mathbf{x}_{\mathbf{1}}+\mathbf{4} \mathbf{x}_{\mathbf{2}}+\mathbf{2} \mathbf{x}_{\mathbf{3}}$ | $\mathbf{F}$ |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 2 | 1 |
| 0 | 1 | 0 | 4 | 1 |
| 0 | 1 | 1 | 6 | 1 |
| 1 | 0 | 0 | -2 | 0 |
| 1 | 0 | 1 | 0 | 0 |
| 1 | 1 | 0 | 2 | 1 |
| 1 | 1 | 1 | 4 | 1 |

From the input - output relation is given in the table, the Boolean expression for the output is

$$
\mathrm{F}=\sum \mathrm{m}(1,2,3,6,7)
$$

The K-map for F is


## INTRODUCTION:-

- Switching circuits are also called logic circuits, gates circuits and digital circuits.
- Switching algebra is also called Boolean algebra.
- Boolean algebra is a system of mathematical logic. It is an algebraic system consisting of the set of elements $(0,1)$, two binary operators called OR and AND and unary operator called NOT.
- It is the basic mathematical tool in the analysis and synthesis of switching circuits.
- It is a way to express logic functions algebraically.
- Any complex logic can be expressed by a Boolean function.
- The Boolean algebra is governed by certain well developed rules and laws.


## AXIOMS AND LAWS OF BOOLEAN ALGEBRA:-

Axioms or postulates of Boolean algebra are set of logical expressions that are accepted without proof and upon which we can build a set of useful theorems. Actually, axioms are nothing more than the definitions of the three basic logic operations AND, OR and INVERTER. Each axiom can be interpreted as the outcome of an operation performed by a logic gate.

## AND operation

Axiom 1: $0.0=0$

Axiom 2: $0.1=0$
Axiom 3: $1.0=0$
Axiom 2: 1. $1=1$

## OR operation

Axiom 5: $0+0=$
0
Axiom 6: $0+1=1$
Axiom 7: $1+0=1$
Axiom 8: $1+1=1$

## NOT operation

Axiom 9: $\overline{1}=0$
Axiom 10:0 $=1$

## 1. Complementation Laws:-

The term complement simply means to invert, i.e. to changes 0 s to 1 s and 1 s to 0 s . The five laws of complementation are as follows:

Law 1: 0 三 1
Law 2: $\overline{1}=0$
Law 3: if $\mathrm{A}=0$, then $\mathrm{A}=$
1 Law 4: if $\underline{A}=1$, then $\mathrm{A}=$
0
Law 5: $\overline{\mathrm{A}}=0$ (double complementation law)

## 2. OR Laws:-

The four OR laws are as follows
Law 1: $\mathrm{A}+0=0$ (Null law)
Law 2: $\mathrm{A}+1=1$ (Identity law)
Law 3: $\mathrm{A}+\mathrm{A}=\mathrm{A}$
Law 4: $A+\bar{A}=1$

## 3. AND Laws:-

The four AND laws are as follows
Law 1: A. $0=0$ (Null law)
Law 2: A. $1=1$ (Identity
law) Law 3: A. A = A
Law 4: A. $\overline{\mathrm{A}}=0$

## 4. Commutative Laws:-

Commutative laws allow change in position of AND or OR variables. There are two commutative laws.
Law 1: $\mathrm{A}+\mathrm{B}=\mathrm{B}+\mathrm{A}$
Proof

| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{A}+\mathbf{B}$ |
| :--- | :--- | :--- |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 1 |


$=\quad$| $\mathbf{B}$ | $\mathbf{A}$ | $\mathbf{B}+\mathbf{A}$ |
| :--- | :--- | :--- |
| 0 | 0 | 0 |
| 0 | 1 | 1 |
| 1 | 0 | 1 |
| 1 | 1 | 1 |

Law 2: $\mathrm{A} . \mathrm{B}=\mathrm{B} . \mathrm{A}$

## Proof

| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{A . B}$ |
| :--- | :--- | :--- |
| 0 | 0 | 0 |
| 0 | 1 | 0 |
| 1 | 0 | 0 |
| 1 | 1 | 1 |$\quad=$| $\mathbf{B}$ | $\mathbf{A}$ | B. $\mathbf{A}$ |
| :--- | :--- | :--- |
| 0 | 0 | 0 |
| 0 | 1 | 0 |
| 1 | 0 | 0 |
| 1 | 1 | 1 |

This law can be extended to any number of variables. For example
A.B. $\mathrm{C}=\mathrm{B} . \mathrm{C} \cdot \mathrm{A}=\mathrm{C} . \mathrm{A} . \mathrm{B}=\mathrm{B} . \mathrm{A} . \mathrm{C}$

## 5. Associative Laws:-

The associative laws allow grouping of variables. There are 2 associative laws.
Law 1: $(A+B)+C=A+(B+C)$

## Proof

| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{C}$ | $\mathbf{A}+\mathbf{B}$ | $(\mathbf{A}+\mathbf{B})+\mathbf{C}$ |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 1 |
| 0 | 1 | 0 | 1 | 1 |
| 0 | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 1 | 1 |
| 1 | 0 | 1 | 1 | 1 |
| 1 | 1 | 0 | 1 | 1 |
| 1 | 1 | 1 | 1 | 1 |


| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{C}$ | $\mathbf{B}+\mathbf{C}$ | $\mathbf{A + ( B + C )}$ |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 1 |
| 0 | 1 | 0 | 1 | 1 |
| 0 | 1 | 1 | 1 | 1 |
| 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 | 1 |
| 1 | 1 | 0 | 1 | 1 |
| 1 | 1 | 1 | 1 | 1 |

Law 2: (A .B) $\mathrm{C}=\mathrm{A}(\mathrm{B} . \mathrm{C})$

## Proof

| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{C}$ | $\mathbf{A B}$ | $(\mathbf{A B}) \mathbf{C}$ |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 0 |
| 0 | 1 | 0 | 0 | 0 |
| 0 | 1 | 1 | 0 | 0 |
| 1 | 0 | 0 | 0 | 0 |
| 1 | 0 | 1 | 0 | 0 |
| 1 | 1 | 0 | 1 | 0 |
| 1 | 1 | 1 | 1 | 1 |


| A | B | C | B.C | A(B.C) |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 0 |
| 0 | 1 | 0 | 0 | 0 |
| 0 | 1 | 1 | 1 | 0 |
| 1 | 0 | 0 | 0 | 0 |
| 1 | 0 | 1 | 0 | 0 |
| 1 | 1 | 0 | 0 | 0 |
| 1 | 1 | 1 | 1 | 1 |

This law can be extended to any number of variables. For example
$\mathrm{A}(\mathrm{BCD})=(\mathrm{ABC}) \mathrm{D}=(\mathrm{AB})(\mathrm{CD})$

## 6. Distributive Laws:-

The distributive laws allow factoring or multiplying out of expressions. There are two distributive laws.
Law 1: $A(B+C)=A B+A C$
Proof

| $A$ | $B$ | $C$ | $B+C$ | $A(B+C)$ |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 0 |
| 0 | 1 | 0 | 1 | 0 |
| 0 | 1 | 1 | 1 | 0 |
| 1 | 0 | 0 | 0 | 0 |
| 1 | 0 | 1 | 1 | 1 |
| 1 | 1 | 0 | 1 | 1 |
| 1 | 1 | 1 | 1 | 1 |

Law 2: $\mathrm{A}+\mathrm{BC}=(\mathrm{A}+\mathrm{B})$
$(\mathrm{A}+\mathrm{C})$ Proof $\mathrm{RHS}=(\mathrm{A}+\mathrm{B})(\mathrm{A}+\mathrm{C})$

$$
\begin{aligned}
& =\mathrm{AA}+\mathrm{AC}+\mathrm{BA}+\mathrm{BC} \\
& =\mathrm{A}+\mathrm{AC}+\mathrm{AB}+\mathrm{BC} \\
& =\mathrm{A}(1+\mathrm{C}+\mathrm{B})+\mathrm{BC} \\
& =\mathrm{A} \cdot 1+\mathrm{BC} \\
& =\mathrm{A}+\mathrm{BC} \\
& =\mathbf{L H S}
\end{aligned}
$$

## 7. Redundant Literal Rule

(RLR):- Law 1: $\mathrm{A}+\mathrm{AB}=$
A + B

## Proof

$$
\begin{aligned}
\mathrm{A}+-\overline{\mathrm{AB}} & =(\mathrm{A}+-\mathrm{A})(\mathrm{A}+\mathrm{B}) \\
& =1 \cdot(\mathrm{~A}+\mathrm{B}) \\
& =\mathrm{A}+\mathrm{B}
\end{aligned}
$$

Proof
Law 2: $\mathrm{A}(\overline{\mathrm{A}}+\mathrm{B})=$

$$
\mathrm{AB} \overline{\mathrm{~A}}(\mathrm{~A}+\mathrm{B} \overline{)}=\mathrm{AA}+
$$

AB

$$
\begin{aligned}
& =0+\mathrm{AB} \\
& =\mathrm{AB}
\end{aligned}
$$

## 8. Idempotence Laws:-

Idempotence means same value.
Law 1: A. A = A
Proof
If $\mathrm{A}=0$, then $\mathrm{A} . \mathrm{A}=0.0=0=$
A If $\mathrm{A}=1$, then $\mathrm{A} . \mathrm{A}=1.1=1$
= A
This law states that AND of a variable with itself is equal to that variable only.

Law 2: $\mathrm{A}+\mathrm{A}=\mathrm{A}$
Proof
If $\mathrm{A}=0$, then $\mathrm{A}+\mathrm{A}=0+0=0=$
A If $\mathrm{A}=1$, then $\mathrm{A}+\mathrm{A}=1+1=1$
= A
This law states that OR of a variable with itself is equal to that variable only.

## 9. Absorption Laws:-

There are two laws:

$$
\text { Law 1: } A+A \cdot B=A
$$

Proof

$$
A+A \cdot B=A(1+B)=A \cdot 1=A
$$

| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{A B}$ | $\mathbf{A}+\mathbf{A B}$ |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 |
| 0 | 1 | 0 | 0 |
| 1 | 0 | 0 | 1 |
| 1 | 1 | 1 | 1 |

Law 2: $\mathrm{A}(\mathrm{A}+\mathrm{B})=\mathrm{A}$
Proof

$$
\mathrm{A}(\mathrm{~A}+\mathrm{B})=\mathrm{A} \cdot \mathrm{~A}+\mathrm{A} \cdot \mathrm{~B}=\mathrm{A}+\mathrm{AB}=\mathrm{A}(1+\mathrm{B})=\mathrm{A} \cdot 1=\mathrm{A}
$$

| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{A}+\mathbf{B}$ | $\mathbf{A}(\mathbf{A}+\mathbf{B})$ |
| :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 |
| 0 | 1 | 1 | 0 |


| 1 | 0 | 1 | 1 |
| :--- | :--- | :--- | :--- |
| 1 | 1 | 1 | 1 |

10. Consensus Theorem (Included Factor Theorem):-

Theorem 1:

$$
\mathrm{AB}+\mathrm{AC}+\mathrm{BC}=\mathrm{AB}+\mathrm{AC}
$$

Proof

$$
\begin{aligned}
\mathbf{L H S} & =\mathrm{AB}+\underline{\mathrm{A}} \mathrm{C}+\mathrm{BC} \\
& =\mathrm{AB}+\overline{\mathrm{AC}}+\mathrm{BC}(\mathrm{~A}+\overline{\mathrm{A}}) \\
& =\mathrm{AB}+\overline{\mathrm{AC}}+\mathrm{BCA}+\mathrm{B} \overline{\mathrm{C}} \mathrm{~A} \\
& =\mathrm{AB}(1+-\mathrm{C})+\mathrm{AC}(1+\mathrm{B}) \\
& =\mathrm{AB}(1)+\mathrm{AC}(1)
\end{aligned}
$$

$\begin{array}{ll}\text { Theorem } & =\mathrm{AB}+\overline{\mathrm{AC}} \\ \text { 2: } & =\text { RHS }\end{array}$
$=$ RHS
Proof $(\mathrm{A}+\mathrm{B}) \overline{(\mathrm{A}}+\mathrm{C})(\mathrm{B}+\mathrm{C})=(\mathrm{A}+\mathrm{B}) \overline{(\mathrm{A}}+\mathrm{C})$

$$
\begin{aligned}
\mathrm{LHS} & =(\mathrm{A}+\mathrm{B}) \overline{(\mathrm{A}}+\overline{\mathrm{C}})(\mathrm{B}+\mathrm{C}) \\
& =(\mathrm{AA}+\mathrm{AC}+\overline{\mathrm{BA}}+\overline{\mathrm{BC}})(\mathrm{B}+\mathrm{C}) \\
& =(\mathrm{AC}+\mathrm{BC}+\overline{\mathrm{AB}})(\mathrm{B}+\mathrm{C}) \\
& =\mathrm{ABC}+\mathrm{BC}+\overline{\mathrm{AB}}+\mathrm{AC}+\mathrm{BC}+\mathrm{ABC} \\
& =\mathrm{AC}+\mathrm{BC}+\overline{\mathrm{AB}}
\end{aligned}
$$

$$
\begin{aligned}
\mathrm{RHS} & =(\overline{\mathrm{A}}+\mathrm{B}) \overline{(\mathrm{A}}+\mathrm{C}) \\
& =\overline{\mathrm{AA}}+\mathrm{AC}+\mathrm{BC}+\mathrm{AB} \\
& =\mathrm{AC}+\mathrm{BC}+\overline{\mathrm{AB}} \\
& =\mathrm{LHS}
\end{aligned}
$$

## 11. Transposition Theorem:-

Theorem:

$$
\mathrm{AB}+\overline{\mathrm{A}} \mathrm{C}=(\mathrm{A}+\mathrm{C}) \overline{(\mathrm{A}}+\mathrm{B})
$$

Proof

$$
\begin{aligned}
\mathbf{R H S} & =(\mathrm{A}+\mathrm{C}) \overline{(\mathrm{A}}+\mathrm{B}) \\
& =\overline{\mathrm{A}} \overline{\mathrm{~A}}+\overline{\mathrm{CA}}+\mathrm{AB}+\mathrm{CB} \\
& =\underline{0}+\overline{\mathrm{A}}+\mathrm{AB}+\mathrm{BC}- \\
& =\mathrm{AC}+\mathrm{AB}+\mathrm{BC}(\overline{\mathrm{~A}+\mathrm{A})} \\
& =\mathrm{AB}+\underline{\mathrm{ABC}}+\mathrm{AC}+\mathrm{ABC} \\
& =\mathrm{AB}+\overline{\mathrm{A}} \mathrm{C} \\
& =\mathrm{LHS}
\end{aligned}
$$

## 12. De Morgan's Theorem:-

De Morgan's theorem represents two laws in Boolean algebra.

Law 1: $\mathrm{A}+\mathrm{B}=\mathrm{A} \cdot \mathrm{B}$

Proof

| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{A}+\mathbf{B}$ | $\overline{\mathbf{A}+\mathbf{B}}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 1 |
| 0 | 1 | 1 | 0 |


| $\mathbf{A}$ | $\mathbf{B}$ | $\overline{\mathbf{A}}$ | $\overline{\mathbf{B}}$ | $\overline{\mathbf{A}} \overline{\mathbf{B}}$ |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 1 | 1 | 1 |
| 0 | 1 | 1 | 0 | 0 |

This law states that the complement of a sum of variables is equal to the product of their individual complements.
Law 2: $A \cdot B=A+B$
Proof

| $\mathbf{A}$ | $\mathbf{B}$ | $\mathbf{A . B}$ | $\mathbf{A . B}$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 1 |
| 0 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 |
| 1 | 1 | 1 | 0 |


$=$| $\mathbf{A}$ | $\mathbf{B}$ | $\overline{\mathbf{A}}$ | $\overline{\mathbf{B}}$ | $\mathbf{A}+\overline{\mathbf{B}}$ |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 1 | 1 | 1 |
| 0 | 1 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 | 1 |
| 1 | 1 | 0 | 0 | 0 |

This law states that the complement of a product of variables is equal to the sum of their individual complements.

## DUALITY:-

The implication of the duality concept is that once a theorem or statement is proved, the dual also thus stand proved. This is called the principle of duality.

$$
\left[\mathrm{f}(\mathrm{~A}, \mathrm{~B}, \mathrm{C}, \ldots ., 0,1,+, \cdot \cdot]_{\mathrm{d}}=\mathrm{f}(\mathrm{~A}, \mathrm{~B}, \mathrm{C}, \ldots ., 1,0, \cdot,+)\right.
$$

Relations between complement and dual

$$
\mathrm{f}_{\mathrm{c}}(\mathrm{~A}, \mathrm{~B}, \mathrm{C}, \ldots . .)=\mathrm{f}(\mathrm{~A}, \mathrm{~B}, \mathrm{C}, \ldots . .)=\overline{\mathrm{f}_{\mathrm{d}}}(\overline{\mathrm{~A}}, \mathrm{~B}, \overline{\mathrm{C}}, \ldots)
$$

$$
f_{d}(A, B, C, \ldots . .)=f(\overline{(A, B}, \bar{B}, \ldots)=\overline{f_{c}}(\bar{A}, \bar{B}, C, \ldots . .)
$$

The first relation states that the complement of a function $f(A, B, C, \ldots)$ can be obtained by complementing all the variables in the dual function $\mathrm{f}_{\mathrm{d}}(\mathrm{A}, \mathrm{B}, \mathrm{C}, \ldots .$.$) .$
The second relation states that the dual can be obtained by complementing all the literals in $f$ (A, B, C, ....).

## DUALS:-

Given expression
Dual

1. $0=1$
2. $0 \cdot 1=0$
3. $0 \cdot 0=0$
4. $1 \cdot 1=1$
5. $\mathrm{A} \cdot 0=0$
6. $\mathrm{A} \cdot 1=\mathrm{A}$
7. $\mathrm{A} \cdot \mathrm{A}=\mathrm{A}$
8. $A \cdot A=0$
9. $\mathrm{A} \cdot \mathrm{B}=\mathrm{B} \cdot \mathrm{A}$
10. $A \cdot(B \cdot C)=(A \cdot B) \cdot C$
11. $A \cdot(B+C)=A B+A C$
12. $A(A+B)=A$
13. $\underline{A}:(\underline{A} \cdot B)=A \cdot B$
$1=0$
$1+0=1$
$1+1=1$
$0+0=0$
$\mathrm{A}+1=1$
$\mathrm{A}+0=\mathrm{A}$
$\mathrm{A}+\mathrm{A}=\mathrm{A}$
$\mathrm{A}+\mathrm{A}=1$
$\mathrm{A}+\mathrm{B}=\mathrm{B}+\mathrm{A}$
$A+(B+C)=(A+B)+C$
$A+B C=(A+B)(A+C)$
$\mathrm{A}+\mathrm{AB}=\mathrm{A}$
$\underline{A+A}+\underline{B} \equiv A+B$
14. $A B=A+B$
15. $(A+B)(A+C)(B+C)=(A+B)(A+C)$
$A+B=A B$
16. $\mathrm{A}+\mathrm{BC}=(\mathrm{A}+\mathrm{B})(\mathrm{A}+\mathrm{C})$
17. $(A+C)(A+B)=A B+A C$
18. $(\mathrm{A}+\mathrm{B})(\mathrm{C}+\mathrm{D})=\mathrm{AC}+\mathrm{AD}+\mathrm{BC}+\mathrm{BD}$
19. $\overline{\mathrm{A}}+\overline{\mathrm{B}}=\mathrm{AB}+\mathrm{AB}+\mathrm{AB}$
20. $\mathrm{AB}+\mathrm{A}+\mathrm{AB}=0$
$\mathrm{AB}+{ }^{-} \mathrm{AC}+\mathrm{BC}=\mathrm{AB}+\mathrm{AC}$
$\mathrm{A}(\overline{\mathrm{B}}+\mathrm{C})=\overline{\mathrm{A}} \mathrm{B}+\mathrm{AC}$
$\mathrm{AC}+\mathrm{AB}=(\mathrm{A}+\mathrm{B})(\mathrm{A}+\mathrm{C})$
$(\mathrm{AB}+\mathrm{CD})=(\mathrm{A}+\mathrm{C})(\underline{\mathrm{A}}+\mathrm{D})(\mathrm{B}+\mathrm{C})(\mathrm{B}+\mathrm{D})$
$\overline{\mathrm{AB}}=(\overline{\mathrm{A}}+\mathrm{B})(\mathrm{A}+\mathrm{B})(\mathrm{A}+\mathrm{B})$
$A+B \cdot A \cdot(A+B)=1$

## SUM - OF - PRODUCTS FORM:-

- This is also called disjunctive Canonical Form (DCF) or Expanded Sum of Products Form or Canonical Sum of Products Form.
- In this form, the function is the sum of a number of products terms where each product term contains all variables of the function either in complemented or uncomplemented form.
- This can also be derived from the truth table by finding the sum of all the terms that corresponds to those combinations for which ' f ' assumes the value 1 .
For example

$$
\begin{aligned}
\mathrm{f}(\mathrm{~A}, \mathrm{~B}, \mathrm{C}) & =\overline{\mathrm{AB}}+\overline{\mathrm{BC}} \\
& =\overline{\mathrm{A} B}(\mathrm{C}+\overline{\mathrm{C}})+\overline{\mathrm{B}} \mathrm{C}(\mathrm{~A}+\mathrm{A}) \\
& =\overline{\mathrm{A}} \overline{\mathrm{BC}}+\overline{\mathrm{ABC}}+\mathrm{ABC}+\overline{\mathrm{A}} \mathrm{BC}
\end{aligned}
$$

- The product term which contains all the variables of the functions either in complemented or uncomplemented form is called a minterm.
- The minterm is denoted as mo, $\mathrm{ml}, \mathrm{m} 2 \ldots$.
- An ' $n$ ' variable function can have $2 n$ minterms.
- Another way of representing the function in canonical SOP form is the showing the sum of minterms for which the function equals to 1 .
For example

$$
\begin{aligned}
& \mathrm{f}(\mathrm{~A}, \mathrm{~B}, \mathrm{C})=\mathrm{m}_{1}+\mathrm{m}_{2}+\mathrm{m}_{3}+\mathrm{m}_{5} \\
& \text { or } \\
& \mathrm{f}(\mathrm{~A}, \mathrm{~B}, \mathrm{C})=\sum \mathrm{m}(1,2,3,5)
\end{aligned}
$$

where $\sum \mathrm{m}$ represents the sum of all the minterms whose decimal codes are given the parenthesis.

## PRODUCT- OF - SUMS FORM:-

- This form is also called as Conjunctive Canonical Form (CCF) or Expanded Product - of - Sums Form or Canonical Product Of Sums Form.
- This is by considering the combinations for which $\mathrm{f}=0$
- Each term is a sum of all the variables.
- The function $\mathrm{f}(\mathrm{A}, \mathrm{B}, \mathrm{C})=(\overline{\mathrm{A}}+\underline{\mathrm{B}}+\overline{\mathrm{C}} \cdot \mathrm{C})+(\underline{\mathrm{A}}+\underline{\mathrm{B}}+\overline{\mathrm{C}} \cdot \mathrm{C})$

$$
=(\bar{A}+\mathrm{B}+\mathrm{C})(\mathrm{A}+\mathrm{B}+\mathrm{C})(\mathrm{A}+\mathrm{B}+\mathrm{C})(\mathrm{A}+\mathrm{B}+\mathrm{C})
$$

- The sum term which contains each of the ' $n$ ' variables in either complemented or uncomplemented form is called a maxterm.
- Maxterm is represented as $\mathrm{M}_{0}, \mathrm{M}_{1}, \mathrm{M}_{2}, \ldots \ldots$.

Thus CCF of ' $f$ ' may be written as

$$
f(A, B, C)=M_{0} \cdot M_{4} \cdot M_{6}
$$

$\mathrm{M}_{7}$
or
$\mathrm{f}(\mathrm{A}, \mathrm{B}, \mathrm{C})=(0,4,6,7)$
Where represented the product of all maxterms.

## CONVERSION BETWEEN CANONICAL FORM:-

The complement of a function expressed as the sum of minterms equals the sum of minterms missing from the original function.

Example:-
$f(A, B, C)=\sum m(0,2,4,6,7)$
This has a complement that can be expressed as
$\mathrm{f}(\mathrm{A}, \mathrm{B}, \mathrm{C})=\sum \mathrm{m}(1,3,5)=\mathrm{m}_{1}+\mathrm{m}_{3}+\mathrm{m}_{5}$
If we complement $f$ by De- Morgan's theorem we obtain ' $f$ ' in a form. $\mathrm{f}=\left(\mathrm{m}_{1}+\mathrm{m}_{3}+\mathrm{m} \overline{)}\right)=\overline{\mathrm{m}}_{1} \cdot \overline{\mathrm{~m}}_{3} . \mathrm{m}_{5}$

$$
=\mathrm{M}_{1} \mathrm{M}_{3} \mathrm{M}_{5}=\prod \mathrm{M}(1,3,5)
$$

## Example:-

Expand $\mathbf{A}(\bar{A}+B)(\bar{A}+B+\bar{C})$ to maxterms and minterms.

## Solution:-

In POS form

$$
\begin{aligned}
& \mathrm{A}(\underline{\mathrm{~A}}+\mathrm{B}) \overline{(\mathrm{A}}+\mathrm{B}+ \\
& \text { C) } \mathrm{A}=\mathrm{A}+\mathrm{BB}+\mathrm{C} \underline{C} \\
& =(\mathrm{A}+\mathrm{B})(\mathrm{A} \pm \mathrm{B})+\mathrm{C} \cdot \mathrm{C} \\
& =(\mathrm{A}+\mathrm{B}+\mathrm{CC})(\mathrm{A}+\mathrm{B}+\overline{\mathrm{CC}}) \\
& =(A+B+C)(A+B \overline{+C})(A \overline{+} B+C)(A+B \overline{+} \\
& \text { C) } \mathrm{A}+\mathrm{B}=\mathrm{A}+\mathrm{B}+\mathrm{C} \cdot \mathrm{C} \\
& =\overline{(A}+B+C) \overline{(A+B+\bar{C})}
\end{aligned}
$$

Therefore

$$
\begin{aligned}
&\mathrm{A}(\overline{\mathrm{~A}}+\mathrm{B}) \overline{(\mathrm{A}}+\mathrm{B}+\overline{\mathrm{C}}) \\
&=(\mathrm{A}+\mathrm{B}+\mathrm{C})(\mathrm{A}+\mathrm{B}+\mathrm{C})(\mathrm{A}+\mathrm{B}+\mathrm{C})(\mathrm{A}+\mathrm{B}+\overline{\mathrm{C}}) \overline{(\mathrm{A}+\mathrm{B}+\mathrm{C})} \overline{(\mathrm{A}+\mathrm{B}+\mathrm{C})} \\
&=(000)(001)(010)(011)(100)(101) \\
&=\mathrm{M}_{0} \cdot \mathrm{M}_{1} \cdot \mathrm{M}_{2} \cdot \mathrm{M}_{3} \cdot \mathrm{M}_{4} \cdot \mathrm{M}_{5} \\
&=\prod \mathrm{M}(0,1,2,3,4,5)
\end{aligned}
$$

The maxterms $\mathrm{M}_{6}$ and $\mathrm{M}_{7}$ are missing in the POS form.
So, the SOP form will contain the minterms 6 and 7

## KARNAUGH MAP OR K-MAP:-

- The K- map is a chart or a graph, composed of an arrangement of adjacent cells, each representing a particular combination of variables in sum or product form.
- The K- map is systematic method of simplifying the Boolean expression.


## TWO VARIABLE K-MAP:-

A two variable expression can have $2^{2}=4$ possible combinations of the input variables A and B .

## Mapping of SOP Expression:-

- The 2 variable K-map has $2^{2}=4$ squares. These squares are called cells.
- A ' 1 ' is placed in any square indicates that corresponding minterm is included in the output expression, and a 0 or no entry in any square indicates that the corresponding minterm does not appear in the expression for output.

B


Example:-
Map expression $f=\overline{\mathbf{A}} \mathbf{B}+\mathbf{A} \overline{\mathbf{B}}$

## Solution:-

The expression minterms is

$$
\mathrm{F}=\mathrm{m}_{1}+\mathrm{m}_{2}=\mathrm{m}(1,2)
$$

0 | 0 | 1 |  |  |
| :--- | :--- | :--- | :--- |
|  | 0 | 0 |  |
|  | 1 |  |  |
| 1 | 2 | 0 | 3 |

## Minimization of SOP Expression:-

To minimize a Boolean expression given in the SOP form by using K- map, the adjacent squares having 1s, that is minterms adjacent to each other are combined to form larger squares to eliminate some variables.
The possible minterm grouping in a two variable K- map are shown below


- Two minterms, which are adjacent to each other, can be combined to form a bigger square called 2 square or a pair. This eliminates one variable that is not common to both the minterms.
- Two 2-squares adjacent to each other can be combined to form a 4- square. A 4- square eliminates 2 variables. A 4 -square is called a quad.
- Consider only those variables which remain constant throughout the square, and ignore the variables which are varying. The non-complemented variable is the variable remaining constant as 1.The complemented variable is the variable remaining constant as a 0 and the variables are written as a product term.


## Example:-

Reduce the expression $\mathrm{f}=\overline{\mathrm{AB}}+\mathrm{AB}+\mathrm{AB}$ using mapping.

## Solution:-

Expressed in terms of minterms, the given expression is

$$
\mathrm{f}=\mathrm{m}_{0}+\mathrm{m}_{1}+\mathrm{m}_{3}=\sum \mathrm{m}(0,1,3)
$$


$\mathbf{F}=\mathbf{A}+\mathbf{B}$

## Mapping of POS Expression:-

Each sum term in the standard POS expression is called a Maxterm. A function in two variables (A,B) has 4 possible maxterms, $A+B, A+B, A+B$ and $A+B$. They are represented as $M_{0}, M_{1}, M_{2}$ and $M_{3}$ respectively.


The maxterm of a two variable K -
map Example:-
Plot the expression $f=(\mathbf{A}+\mathbf{B})(\overline{\mathbf{A}}+\underline{B})(\overline{\mathbf{A}}+-$

## B) Solution:-

Expression interms of maxterms is $\mathrm{f}=\pi \mathrm{M}(0,2,3)$


## Minimization of POS Expressions:-

In POS form the adjacent 0 s are combined into large square as possible. If the squares having complemented variable then the value remain constant as a 1 and the non-complemented variable if its value remains constant as a 0 along the entire square and then their sum term is written.
The possible maxterms grouping in a two variable K-map are shown below


## Example:-

Reduce the expression $f=(A+B)(\bar{A}+\bar{B})(A+B)$ using mapping

## Solution:-

The given expression in terms of maxterms is $\mathrm{f}=\pi \mathrm{M}(0,1,3)$


## THREE VARIABLE K-MAP:-

A function in three variables ( $\mathrm{A}, \mathrm{B}, \mathrm{C}$ ) can be expressed in SOP and POS form having eight possible combination. A three variable K - map have 8 squares or cells and each square on the map represents a minterm or maxterm is shown in the figure below.

|  | 00 | 01 | 11 | 10 |
| :---: | :---: | :---: | :---: | :---: |
| 0 | $\begin{aligned} & \overline{\overline{A B} \bar{C}}{ }^{0} \\ & \left(m_{0}\right) \end{aligned}$ | $\overline{\bar{A} \bar{B} C^{1}}$ $\left(m_{1}\right)$ | $\overline{A B C}{ }^{3}$ $\left(m_{3}\right)$ | $\begin{aligned} & \overline{A B} \bar{C}^{2} \\ & \left(\mathrm{~m}_{2}\right) \end{aligned}$ |
| 1 | $\begin{aligned} & \mathrm{AB} \overline{\mathrm{~B}}^{4} \\ & \left(\mathrm{~m}_{4}\right) \end{aligned}$ | $\begin{aligned} & A \bar{B} C^{5} \\ & \left(m_{5}\right) \end{aligned}$ | $\begin{aligned} & \mathrm{ABC}^{7} \\ & \left(\mathrm{~m}_{7}\right) \end{aligned}$ | $\begin{aligned} & A B \bar{C}^{6} \\ & \left(m_{6}\right) \end{aligned}$ |

(a) Minterms

(b) Maxterms

## Example:-

Map the expression $f=\overline{\mathbf{A B}} \overline{\mathbf{C}}+\mathrm{A} \overline{\mathbf{B C}}+\overline{\mathbf{A}} \mathbf{B} \overline{\mathbf{C}}+\mathbf{A B} \overline{\mathbf{C}}$

## +ABC Solution:-

So in the SOP form the expression is $\mathrm{f}=\sum \mathrm{m}(1,5,2,6,7)$

A | 00 | 01 | 11 | 10 |
| :---: | :---: | :---: | :---: |
| 0 | 0 | $1^{0}$ | $0^{3}$ |
| 1 | $0^{2}$ | $1^{5}$ | $1^{7}$ |

Example:-
Map the expression $\mathbf{f}=(\mathbf{A}+\mathbf{B}+\mathbf{C})(\overline{\mathbf{A}}+\mathbf{B}+\overline{\mathrm{C}})(\overline{\mathbf{A}}+\overline{\mathbf{B}}+\overline{\mathbf{C}})(\mathbf{A}+\overline{\mathbf{B}}+\overline{\mathbf{C}})(\overline{\mathbf{A}}+\overline{\mathbf{B}}+$
C) Solution:-

So in the POS form the expression is $\mathrm{f}=\pi \mathrm{M}(0,5,7,3,6)$

| A ${ }^{\text {B }}$ | 00 | 01 | 11 | 10 |
| :---: | :---: | :---: | :---: | :---: |
| 0 |  |  | $0^{3}$ | $1^{2}$ |
| 1 | $1{ }^{4}$ | $0^{5}$ | $0^{7}$ | $0^{6}$ |

## Minimization of SOP and POS Expressions:-

For reducing the Boolean expressions in SOP (POS) form the following steps are given below

- Draw the K-map and place $1 \mathrm{~s}(0 \mathrm{~s})$ corresponding to the minterms (maxterms) of the SOP (POS) expression.
- In the map $1 \mathrm{~s}(0 \mathrm{~s})$ which are not adjacent to any other $1(0)$ are the isolated minterms (maxterms). They are to be read as they are because they cannot be combined even into a 2 -square.
- For those $1 \mathrm{~s}(0 \mathrm{~s})$ which are adjacent to only one other $1(0)$ make them pairs ( 2 squares).
- For quads (4- squares) and octet ( 8 squares) of adjacent $1 \mathrm{~s}(0 \mathrm{~s}$ ) even if they contain some $1 \mathrm{~s}(0 \mathrm{~s})$ which have already been combined. They must geometrically form a square or a rectangle.
- For any $1 \mathrm{~s}(0 \mathrm{~s})$ that have not been combined yet then combine them into bigger squares if possible.
- Form the minimal expression by summing (multiplying) the product (sum) terms of all the groups.

Some of the possible combinations of minterms in SOP form


These possible combinations are also for POS but 1s are replaced by 0s.

## FOUR VARIABLE K-MAP:-

A four variable (A, B, C, D) expression can have $2^{4}=16$ possible combinations of input variables. A four variable K-map has $2^{4}=16$ squares or cells and each square on the map represents either a minterm or a maxterm as shown in the figure below. The binary number designations of the rows and columns are in the gray code. The binary numbers along the top of the map indicate the conditions of C and D along any column and binary numbers along left side indicate the conditions of $A$ and $B$ along any row. The numbers in the top right corners of the squares indicate the minterm or maxterm desginations.

## SOP FORM



POS FORM

| $A B$ | 00 | 01 | 11 | 10 |
| :---: | :---: | :---: | :---: | :---: |
| 00 | $\underset{\left(M_{0}\right)}{A+C}+D$ | $A+B+C+\bar{D}^{1}$ <br> $\left(\mathrm{M}_{1}\right)$ | $\underset{\left(M_{3}\right)}{\mathrm{A}+\mathrm{B}+\overline{\mathrm{C}}+\overline{\mathrm{D}}^{3}}$ | $A+B+\bar{C}+D^{2}$ <br> $\left(\mathrm{M}_{2}\right)$ |
| 01 | $\begin{gathered} \mathrm{A}+\overline{\mathrm{B}}+\mathrm{C}+\mathrm{D}^{4} \\ \left(\mathrm{M}_{4}\right) \end{gathered}$ | $\begin{gathered} A+\bar{B}+C+\bar{D}^{5} \\ \left(M_{5}\right) \end{gathered}$ | $\begin{gathered} A+\bar{B}+\bar{C}+\bar{D}^{7} \\ \left(M_{7}\right) \end{gathered}$ | $\begin{gathered} A+\bar{B}+\bar{C}+D^{6} \\ \left(M_{6}\right) \end{gathered}$ |
| 11 | $\begin{gathered} \bar{A}+\bar{B}+C+D^{12} \\ \left(\mathrm{M}_{12}\right) \end{gathered}$ | $\overline{\bar{A}}+\begin{gathered} \bar{B}+C+\bar{D} \\ \left(M_{13}\right) \end{gathered}$ | $\begin{gathered} \left.\overline{\mathrm{A}}+\underset{\mathrm{B}}{\mathrm{~B}}+\overline{\mathrm{C}}+\mathrm{M}_{15}\right) \end{gathered}$ | $\begin{gathered} \overline{\mathrm{A}}+\overline{\mathrm{B}}+\overline{\mathrm{C}}+\mathrm{D}^{14} \\ \left(\mathrm{M}_{14}\right) \end{gathered}$ |
| 10 | $\begin{gathered} \bar{A}+B+C+D^{8} \\ \left(M_{8}\right) \end{gathered}$ | $\overline{\bar{A}}+\underset{\left(M_{9}\right)}{B+C}+\bar{D}^{9}$ | $\overline{\bar{A}+}+\underset{\left(M_{11}\right)}{B+\bar{C}+\bar{D}}$ | $\begin{gathered} \bar{A}+\underset{\left(M_{10}\right)}{ }+\bar{C}+D^{10} \\ x^{10} \end{gathered}$ |

## Minimization of SOP and POS Expressions:-

For reducing the Boolean expressions in SOP (POS) form the following steps are given below

- Draw the K-map and place 1s ( 0 s ) corresponding to the minterms (maxterms) of the SOP (POS) expression.
- In the map $1 \mathrm{~s}(0 \mathrm{~s})$ which are not adjacent to any other $1(0)$ are the isolated minterms (maxterms). They are to be read as they are because they cannot be combined even into a 2 -square.
- For those $1 \mathrm{~s}(0 \mathrm{~s})$ which are adjacent to only one other $1(0)$ make them pairs
- For quads (4- squares) and octet (8 squares) of adjacent 1 s ( 0 s ) even if they contain some $1 \mathrm{~s}(0 \mathrm{~s}$ ) which have already been combined. They must geometrically form a square or a rectangle.
- For any $1 \mathrm{~s}(0 \mathrm{~s})$ that have not been combined yet then combine them into bigger squares if possible.
- Form the minimal expression by summing (multiplying) the product (sum) terms of all the groups. Example:-
Reduce using mapping the expression $f=\sum m(0,1,2,3,5,7,8,9,10,12,13)$
Solution:-
The given expression in POS form is $\mathrm{f}=\pi \mathrm{M}(4,6,11,14,15)$ and in $\operatorname{SOP}$ form $\mathrm{f}=\sum \mathrm{m}(0,1,2,3,5,7,8,9$, $10,12,13)$

(a) SOP K-map

(b) POS K-map

The minimal SOP expression is $f_{\text {min }}=B D+A C+A D$

The minimal POS expression is $\mathrm{f}_{\min }=(\mathrm{A}+\mathrm{B}+\mathrm{D})(\mathrm{A}+\mathrm{C}+\mathrm{D})(\mathrm{A}+\mathrm{B}+\mathrm{C})$

## DON'T CARE COMBINATIONS:-

The combinations for which the values of the expression are not specified are called don't care combinations or optional combinations and such expression stand incompletely specified. The output is a don't care for these invalid combinations. The don't care terms are denoted by d or X. During the process of designing using SOP maps, each don't care is treated as 1 to reduce the map otherwise it is treated as 0 and left alone. During the process of designing using POS maps, each don't care is treated as 0 to reduce the map otherwise it is treated as 1 and left alone.

A standard SOP expression with don't cares can be converted into standard POS form by keeping the don't cares as they are, and the missing minterms of the SOP form are written as the maxterms of the POS form. Similarly, to convert a standard POS expression with don't cares can be converted into standard SOP form by keeping the don't cares as they are, and the missing maxterms of the POS form are written as the minterms of the SOP form.

## Example:-

Reduce the expression $f=\sum m(1,5,6,12,13,14)+d(2,4)$ using $K-m a p$.

## Solution:-

The given expression in SOP form is $\mathrm{f}=\sum \mathrm{m}(1,5,6,12,13,14)+\mathrm{d}(2,4)$
The given expression in POS form is $\mathrm{f}=\pi \mathrm{M}(0,3,7,8,9,10,11,15)+\mathrm{d}(2,4)$

(a) SOP K-map

(b) POS K-map

The minimal of SOP expression is $\mathrm{f}_{\text {min }}=\mathrm{BC}+\mathrm{BD}+\mathrm{ACD}$

The minimal of POS expression is $\mathrm{f}_{\min }=(\mathrm{B}+\mathrm{D})(\mathrm{A}+\mathrm{B})(\mathrm{C}+\mathrm{D})$

## SEOUENTIAL LOGIC CIRCUIT

## SEQUENTIAL CIRCUIT:-

- It is a circuit whose output depends upon the present input, previous output and the sequence in which the inputs are applied.

HOW THE SEQUENTIAL CIRCUIT IS DIFFERENT FROM COMBINATIONAL CIRCUIT? :-

- In combinational circuit output depends upon present input at any instant of time and do not use memory. Hence previous input does not have any effect on the circuit. But sequential circuit has memory and depends upon present input and previous output.
- Sequential circuits are slower than combinational circuits and these sequential circuits are harder to design.

[Block diagram of Sequential Logic Circuit]
- The data stored by the memory element at any given instant of time is called the present state of sequential circuit.


## TYPES:-

Sequential logic circuits (SLC) are classified as
(i) Synchronous SLC
(ii) Asynchronous SLC

- The SLC that are controlled by clock are called synchronous SLC and those which are not controlled by a clock are asynchronous SLC.
- Clock:- A recurring pulse is called a clock.


## FLIP-FLOP AND LATCH:-

- A flip-flop or latch is a circuit that has two stable states and can be used to store information.
- A flip-flop is a binary storage device capable of storing one bit of information. In a stable state, the output of a flip-flop is either 0 or 1.
- Latch is a non-clocked flip-flop and it is the building block for the flip-flop.
- A storage element in digital circuit can maintain a binary state indefinitely until directed by an input signal to switch state.
- Storage element that operate with signal level are called latches and those operate with clock transition are called as flip-flops.
- The circuit can be made to change state by signals applied to one or more control inputs and will have one or two outputs.
- A flip-flop is called so because its output either flips or flops meaning to switch back and forth.
- A flip-flop is also called a bi-stable multi-vibrator as it has two stable states. The input signals which command the flip-flop to change state are called excitations.
- Flip-flops are storage devices and can store 1 or 0 .
- Flip-flops using the clock signal are called clocked flip-flops. Control signals are effective only if they are applied in synchronization with the clock signal.
- Clock-signals may be positive-edge triggered or negative-edge triggered.
- Positive-edge triggered flip-flops are those in which state transitions take place only at positive- going edge of the clock pulse.

- Negative-edge triggered flip-flops are those in which state transition take place only at negative- going edge of the clock pulse.

- Some common type of flip-flops include
a) SR (set-reset) F-F
b) D (data or delay) F-F
c) T (toggle) F-F and
d) JK F-F

SR latch:-

- The SR latch is a circuit with two cross-coupled NOR gates or two cross-coupled NAND gates.
- It has two outputs labeled Q and Q'. Two inputs are there labeled S for set and R foe reset.
- The latch has two useful states. When $\mathrm{Q}=0$ and $\mathrm{Q}^{\prime}=1$ the condition is called reset state and when $\mathrm{Q}=1$ and $\mathrm{Q}^{\prime}=0$ the condition is called set state.
- Normally Q and Q' are complement of each other.
- The figure represents a SR latch with two cross-coupled NOR gates. The circuit has NOR gates and as we know if any one of the input for a NOR gate is HIGH then its output will be LOW and if both the inputs are LOW then only the output will be HIGH.

- Under normal conditions, both inputs of the latch remain at 0 unless the state has to be changed. The application of a momentary 1 to the $S$ input causes the latch to go to the set state. The $S$ input must go back to 0 before any other changes take place, in order to avoid the occurrence of an undefined next state that results from the forbidden input condition.
- The first condition ( $\mathrm{S}=1, \mathrm{R}=0$ ) is the action that must be taken by input S to bring the circuit to the set state. Removing the active input from $S$ leaves the circuit in the same state. After both inputs return to 0 , it is then possible to shift to the reset state by momentary applying a 1 to the R input. The 1 can then be removed from $R$, whereupon the circuit remains in the reset state. When both inputs $S$ and $R$ are equal to

0 , the latch can be in either the set or the reset state, depending on which input was most recently a 1.

- If a 1 is applied to both the S and R inputs of the latch, both outputs go to 0 . This action produces an undefined next state, because the state that results from the input transitions depends on the order in which they return to 0 . It also violates the requirement that outputs be the complement of each other. In normal operation, this condition is avoided by making sure that 1 's are not applied to both inputs simultaneously.
- Truth table for SR latch designed with NOR gates is shown below.

| Input |  | Output |  |  |  | Comment |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| S | R | Q | Q' | $\begin{gathered} \mathrm{QNe} \\ \mathrm{xt} \\ \hline \end{gathered}$ | $\begin{aligned} & \text { Q'Ne } \\ & \text { xt } \end{aligned}$ |  |
| 0 | 0 | 0 | 1 | 0 | 1 | No change |
| 0 | 0 | 1 | 0 | 1 | 0 |  |
| 0 | 1 | 0 | 1 | 0 | 1 | Reset |
| 0 | 1 | 1 | 0 | 0 | 1 |  |
| 1 | 0 | 0 | 1 | 1 | 0 | Set |
| 1 | 0 | 1 | 0 | 1 | 0 |  |
| 1 | 1 | 0 | 1 | X | X | Prohibited |
| 1 | 1 | 1 | 0 | X | X | state |



Symbol for SR NOR Latch

## Racing Condition:-

In case of a $S R$ latch when $S=R=1$ input is given both the output will try to become 0 . This is called Racing condition.

SR latch using NAND gate:-

- The below figure represents a SR latch with two cross-coupled NAND gates. The circuit has NAND gates and as we know if any one of the input for a NAND gate is LOW then its output will be HIGH and if both the inputs are HIGH then only the output will be LOW.
- It operates with both inputs normally at 1 , unless the state of the latch has to be changed. The application of 0 to the $S$ input causes output Q to go to 1 , putting the latch in the set state. When the S input goes back to 1 , the circuit remains in the set state. After both inputs go back to 1 , we are allowed to change the state of the latch by placing a 0 in the R input. This action causes the circuit to go to the reset state and stay there even after both inputs return to 1 .

- The condition that is forbidden for the NAND latch is both inputs being equal to 0 at the same time, an input combination that should be avoided.

In comparing the NAND with the NOR latch, note that the input signals for the NAND require the complement of those values used for the NOR latch. Because the NAND latch requires a 0 signal to change its state, it is sometimes referred to as an S'R' latch. The primes (or, sometimes, bars over the letters) designate the fact that the inputs must be in their complement form to activate the circuit.


The above represents the symbol for inverted SR latch or SR latch using NAND gate.
Truth table for SR latch using NAND gate or Inverted SR latch

| S | R | Qnext | Q'next |
| :--- | :--- | :--- | :--- |
| 0 | 0 | Race | Race |
| 0 | 1 | 0 | 1 (Reset) |
| 1 | 0 | 1 | 0 (Set) |
| 1 | 1 | Q (No change) | Q' (No change) |

## D LATCH:-

- One way to eliminate the undesirable condition of the indeterminate state in the SR latch is to ensure that inputs $S$ and $R$ are never equal to 1 at the same time.

(a) Lofic diagram
- This is done in the D latch. This latch has only two inputs: D (data) and En(enable).
- The D input goes directly to the S input, and its complement is applied to the R input.

(Symbol for D-Latch)
- As long as the enable input is at 0 , the cross-coupled SR latch has both inputs at the 1 level and the circuit can't change state regardless of the value of D.
- The below represents the truth table for the D-latch.

| En | D | Next State of Q |
| :--- | :--- | :--- |
| 0 | X | No change |
| 1 | 0 | $\mathrm{Q}=0 ;$ Reset State |
| 1 | 1 | $\mathrm{Q}=1 ;$ Set State |

- The D input is sampled when $\mathrm{En}=1$. If $\mathrm{D}=1$, the Q output goes to 1 , placing the circuit in the set state. If $\mathrm{D}=0$, output Q goes to 0 , placing the circuit in the reset state. This situation provides a path from input D to the output, and for this reason, the circuit is often called a TRANSPARENT latch.


## TRIGGERING METHODS:-

- The state of a latch or flip-flop is switched by a change in the control input. This momentary change is called a trigger, and the transition it causes is said to trigger the flip-flop.
- Flip-flop circuits are constructed in such a way as to make them operate properly when they are part of a sequential circuit that employs a common clock.
- The problem with the latch is that it responds to a change in the level of a clock pulse. For proper operation of a flip-flop it should be triggered only during a signal transition.
- This can be accomplished by eliminating the feedback path that is inherent in the operation of the sequential circuit using latches. A clock pulse goes through two transitions: from 0 to 1 and the return from 1 to 0 .
- A ways that a latch can be modified to form a flip-flop is to produce a flip-flop that triggers only during a signal transition (from 0 to 1 or from 1 to 0 ) of the synchronizing signal (clock) and is disabled during the rest of the clock pulse.



## JK FLIP-FLOP:-

- The JK flip-flop can be constructed by using basic SR latch and a clock. In this case the outputs Q and Q' are returned back and connected to the inputs of NAND gates.
- This simple JK flip Flop is the most widely used of all the flip-flop designs and is considered to be a universal flip-flop circuit.
- The sequential operation of the JK flip flop is exactly the same as for the previous SR flip-flop with the same "Set" and "Reset" inputs.
- The difference this time is that the "JK flip flop" has no invalid or forbidden input states of the SR Latch even when $S$ and $R$ are both at logic " 1 ".
(The below diagram shows the circuit diagram of a JK flip-flop)

- The JK flip flop is basically a gated SR Flip-flop with the addition of a clock input circuitry that prevents the illegal or invalid output condition that can occur when both inputs $S$ and $R$ are equal to logic level " 1 ".
- Due to this additional clocked input, a JK flip-flop has four possible input combinations, "logic 1",
"logic $0 "$ ", "no change" and "toggle".
- The symbol for a JK flip flop is similar to that of an SR bistable latch except the clock input.

(The above diagram shows the symbol of a JK flip-flop.)
- Both the S and the R inputs of the SR bi-stable have now been replaced by two inputs called the J and K inputs, respectively after its inventor Jack and Kilby. Then this equates to: $\mathrm{J}=\mathrm{S}$ and $\mathrm{K}=\mathrm{R}$.
- The two 2-input NAND gates of the gated SR bi-stable have now been replaced by two 3-input NAND gates with the third input of each gate connected to the outputs at Q and Q '.
- This cross coupling of the SR flip-flop allows the previously invalid condition of $\mathrm{S}=$ " 1 " and $\mathrm{R}=$ " 1 " state to be used to produce a "toggle action" as the two inputs are now interlocked.
- If the circuit is now "SET" the J input is inhibited by the " 0 " status of Q' through the lower NAND gate. If the circuit is "RESET" the K input is inhibited by the " 0 " status of Q through the upper NAND gate. As Q and Q ' are always different we can use them to control the input.
(Truth table for JK flip-flop)

| Input |  | Output |  | Comment |
| :---: | :---: | :---: | :---: | :---: |
| $\mathbf{J}$ | K | Q | Qne <br> xt |  |
| 0 | 0 | 0 | 0 | No change |
| 0 | 0 | 1 | 1 |  |
| 0 | 1 | 0 | 0 |  |
| 0 | 1 | 1 | Set |  |
| 1 | 0 | 0 |  | Toggle |
| 1 | 0 | 1 | 1 |  |
| 1 | 1 | 0 | 1 |  |
| 1 | 1 | 1 | 0 |  |

- When both inputs J and K are equal to logic " 1 ", the JK flip flop toggles.


## T FLIP-FLOP:-

- Toggle flip-flop or commonly known as T flip-flop.
- This flip-flop has the similar operation as that of the JK flip-flop with both the inputs $\mathbf{J}$ and K are shorted i.e. both are given the common input.

- Hence its truth table is same as that of JK flip-flop when $\mathrm{J}=\mathrm{K}=0$ and $\mathrm{J}=\mathrm{K}=1$.So its truth table is as follows.

| T | Q | Qne <br> xt | Comment |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | No change |
|  | 1 | 1 |  |
| 1 | 0 | 1 |  |

## CHARACTERISTIC TABLE:-

- A characteristic table defines the logical properties of a flip-flop by describing its operation in tabular form.
- The next state is defined as a function of the inputs and the present state.
- $\mathrm{Q}(\mathrm{t})$ refers to the present state and $\mathrm{Q}(\mathrm{t}+1)$ is the next.
- Thus, $\mathrm{Q}(\mathrm{t})$ denotes the state of the flip-flop immediately before the clock edge, and $\mathrm{Q}(\mathrm{t}+1)$ denotes the state that results from the clock transition.
- The characteristic table for the JK flip-flop shows that the next state is equal to the present state when inputs $J$ and $K$ are both equal to 0 . This condition can be expressed as $Q(t+1)=Q(t)$, indicating that the clock produces no change of state.

Characteristic Table Of JK Flip-Flop

| J | K | $\mathrm{Q}(\mathrm{t}+1)$ |
| :---: | :---: | :---: |
| 0 | 0 | Q(t) No change |
| 0 | 1 | 0 Reset |
| 1 | 0 | 1 Set |
| 1 | 1 | Q'(t) Complement |

- When $\mathrm{K}=1$ and $\mathrm{J}=0$, the clock resets the flip-flop and $\mathrm{Q}(\mathrm{t}+1)=0$. With $\mathrm{J}=1$ and $\mathrm{K}=0$, the flip-flop sets and $\mathrm{Q}(\mathrm{t}+1)=1$. When both J and K are equal to 1 , the next state changes to the complement of the present state, a transition that can be expressed as $\mathrm{Q}(\mathrm{t}+1)=\mathrm{Q}^{\prime}(\mathrm{t})$.
- The characteristic equation for JK flip-flop is represented as

$$
Q(t+1)=J Q^{\prime}+K^{\prime} Q
$$

## Characteristic Table of D Flip-Flop

| D | $\mathrm{Q}(\mathrm{t}+1)$ |
| :---: | :---: |
| 0 | 0 |
| 1 | 1 |

- The next state of a D flip-flop is dependent only on the D input and is independent of the present state.
- This can be expressed as $\mathrm{Q}(\mathrm{t}+1)=\mathrm{D}$. It means that the next-state value is equal to the value of D . Note that the D flip-flop does not have a "no-change" condition and its characteristic equation is written as $Q(t+1)=D$.


## Characteristic Table of T Flip-Flop

| T | $\mathrm{Q}(\mathrm{t}+1)$ |
| :--- | :--- |
| 0 | $\mathrm{Q}(\mathrm{t}) \quad$ No change |
| 1 | $\mathrm{Q}^{\prime}(\mathrm{t})$ |
|  | Complement |

- The characteristic table of T flip-flop has only two conditions: When $\mathrm{T}=0$, the clock edge does not change the state; when $\mathrm{T}=1$, the clock edge complements the state of the flip-flop and the characteristic equation is

$$
\mathrm{Q}(\mathrm{t}+1)=\mathrm{T} \oplus \mathrm{Q}=\mathrm{T}^{\prime} \mathrm{Q}+\mathrm{TQ} \mathrm{Q}^{\prime}
$$

## MASTER-SLAVE JK FLIP-FLOP:-

- The Master-Slave Flip-Flop is basically two gated SR flip-flops connected together in a series configuration with the slave having an inverted clock pulse.
- The outputs from Q and Q' from the "Slave" flip-flop are fed back to the inputs of the "Master" with the outputs of the "Master" flip flop being connected to the two inputs of the "Slave" flip flop.
- This feedback configuration from the slave's output to the master's input gives the characteristic toggle of the JK flip flop as shown below.

The Master-Slave JK Flip Flop


- The input signals J and K are connected to the gated "master" SR flip flop which "locks" the input condition while the clock (Clk) input is "HIGH" at logic level " 1 ".
- As the clock input of the "slave" flip flop is the inverse (complement) of the "master" clock input, the "slave" SR flip flop does not toggle.
- The outputs from the "master" flip flop are only "seen" by the gated "slave" flip flop when the clock input goes "LOW" to logic level " 0 ".
- When the clock is "LOW", the outputs from the "master" flip flop are latched and any additional changes to its inputs are ignored.
- The gated "slave" flip flop now responds to the state of its inputs passed over by the "master" section.
- Then on the "Low-to-High" transition of the clock pulse the inputs of the "master" flip flop are fed through to the gated inputs of the "slave" flip flop and on the "High-to-Low" transition the same inputs are reflected on the output of the "slave" making this type of flip flop edge or pulse-triggered.
- Then, the circuit accepts input data when the clock signal is "HIGH", and passes the data to the output on the falling-edge of the clock signal.
- In other words, the Master-Slave JK Flip flop is a "Synchronous" device as it only passes data with the timing of the clock signal.


## FLIP-FLOP CONVERSIONS:-

## SR Flip Flop to JK Flip Flop

For this J and K will be given as external inputs to S and R . As shown in the logic diagram below, S and R will be the outputs of the combinational circuit.

The truth tables for the flip flop conversion are given below. The present state is represented by Qp and $\mathrm{Qp}+1$ is the next state to be obtained when the J and K inputs are applied.

For two inputs $\mathbf{J}$ and K , there will be eight possible combinations. For each combination of $\mathrm{J}, \mathrm{K}$ and Qp , the corresponding $\mathrm{Qp}+1$ states are found. $\mathrm{Qp}+1$ simply suggests the future values to be obtained by the JK flip flop after the value of Qp . The table is then completed by writing the values of S and R required to get each $\mathrm{Qp}+1$ from the corresponding Qp. That is, the values of $S$ and $R$ that are required to change the state of the flip flop from Qp to $\mathrm{Qp}+1$ are written.

S-R Flip Flop to J-K Flip Flop

Conversion Table

| $J-K$ Inputs |  | Outputs |  | $\mathrm{S}-\mathrm{R}$ Inputs |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | K | $\mathrm{Q} p$ | $Q p+1$ | S | R |
| 0 | 0 | 0 | 0 | 0 | X |
| 0 | 0 | 1 | 1 | X | 0 |
| 0 | 1 | 0 | 0 | 0 | X |
| 0 | 1 | 1 | 0 | 0 | 1 |
| 1 | 0 | 0 | 1 | 1 | 0 |
| 1 | 0 | 1 | 1 | X | 0 |
| 1 | 1 | 0 | 1 | 1 | 0 |
| 1 | 1 | 1 | 0 | 0 | 1 |

Logic Diagram



## JK Flip Flop to SR Flip Flop

- This will be the reverse process of the above explained conversion. $S$ and $R$ will be the external inputs to J and K . J and K will be the outputs of the combinational circuit. Thus, the values of J and K have to be obtained in terms of S, R and Qp.
- A conversion table is to be written using S, R, Qp, Qp+1, J and K.
- For two inputs, S and R, eight combinations are made. For each combination, the corresponding $\mathrm{Qp}+1$ outputs are found out.
- The outputs for the combinations of $\mathrm{S}=1$ and $\mathrm{R}=1$ are not permitted for an SR flip flop. Thus the outputs are considered invalid and the J and K values are taken as "don't cares".

Conversion Table

| S-R Inputs |  | Outputs |  | J-K Inputs |  |
| :---: | :---: | :---: | :---: | :---: | :---: |
| S | R | Qp | p+1 | J | K |
| 0 | 0 | 0 | 0 | 0 | X |
| 0 | 0 | 1 | 1 | X | 0 |
| 0 | 1 | 0 | 0 | 0 | X |
| 0 | 1 | 1 | 0 | X | 1 |
| 1 | 0 | 0 | 1 | 1 | X |
| 1 | 0 | 1 | 1 | X | 0 |
| 1 | 1 |  | lid |  | care |
| 1 | 1 |  | alid |  | care |

Logic Diagram



## SR Flip Flop to D Flip Flop

- $S$ and $R$ are the actual inputs of the flip flop and $D$ is the external input of the flip flop.
- The four combinations, the logic diagram, conversion table, and the K-map for S and R in terms of D and Qp are shown below.


## S-R Flip Flop to D Flip Flop

Conversion Table

| D Input | $\frac{\text { Outputs }}{}$ |  | $\frac{\mathrm{S}-\mathrm{R} \text { Inputs }}{\mathrm{S}}$ |  |
| :---: | :---: | :---: | :---: | :---: |
|  | 0 | 0 | 0 | X |
| 0 | 1 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 | 0 |
| 1 | 1 | 1 | X | 0 |

K-maps


Logic Diagram


## D Flip Flop to SR Flip Flop

- $D$ is the actual input of the flip flop and $S$ and $R$ are the external inputs. Eight possible combinations are achieved from the external inputs $\mathrm{S}, \mathrm{R}$ and Qp .
- But, since the combination of $\mathrm{S}=1$ and $\mathrm{R}=1$ are invalid, the values of $\mathrm{Qp}+1$ and D are considered as "don't cares".
- The logic diagram showing the conversion from D to SR, and the K-map for D in terms of S, R and Qp are shown below.

Conversion Table

| $\frac{S-R ~ I n p u t s ~}{S}$ |  | $\begin{aligned} & \text { Outputs } \\ & Q P Q D+1 \end{aligned}$ |  | D Input |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 1 |
| 0 | 1 | 0 | 0 | 0 |
| 0 | 1 | 1 | 0 | 0 |
| 1 | 0 | 0 | 1 | 1 |
| 1 | 0 | 1 | 1 | 1 |
| 1 | 1 |  | lid | Don't care |
| 1 | 1 |  |  | Don't care |

K-map

$D=S+\bar{R} Q_{n}$

Logic Diagram


## JK Flip Flop to T Flip Flop:-

- J and K are the actual inputs of the flip flop and T is taken as the external input for conversion
- Four combinations are produced with T and Qp. J and K are expressed in terms of T and Qp.
- The conversion table, K-maps, and the logic diagram are given below.
J-K Flip Flop to T Flip Flop
Conversion Table

| T Input | Outputs |  |  |  |
| :---: | :---: | :---: | :---: | :---: |
| $Q$ | $\frac{3-K}{}$ Inputs |  |  |  |
| 0 | 0 | 0 | 0 | X |
| 0 | 1 | 1 | X | 0 |
| 1 | 0 | 1 | 1 | X |
| 1 | 1 | 0 | X | 1 |



## D Flip Flop to JK Flip Flop:-

- In this conversion, D is the actual input to the flip flop and J and K are the external inputs.
- J, K and Qp make eight possible combinations, as shown in the conversion table below. D is expressed in terms of J, K and Qp.
- The conversion table, the K-map for D in terms of J, K and Qp and the logic diagram showing the conversion from D to JK are given in the figure below.


## Conversion Table

| $\frac{3-K \text { Input }}{3}$ |  | $\frac{2}{C u t p u t s}$ |  | $D \ln p u t$ |
| :---: | :--- | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 1 |
| 0 | 1 | 0 | 0 | 0 |
| 0 | 1 | 1 | 0 | 0 |
| 1 | 0 | 0 | 1 | 1 |
| 1 | 0 | 1 | 1 | 1 |
| 1 | 1 | 0 | 1 | 1 |
| 1 | 1 | 1 | 0 | 0 |

K-map



## JK Flip Flop to D Flip Flop:-

- D is the external input and J and K are the actual inputs of the flip flop. D and Qp make four combinations. J and K are expressed in terms of D and Qp .
- The four combination conversion table, the K-maps for J and K in terms of D and Qp.


## J-K Flip Flop to D Flip Flop

Conversion Table

| D Input | $\frac{\text { Outputs }}{Q 0 Q \Delta+1}$ |  | $\frac{3-K \text { Inputs }}{3}$ |  |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | X |
| 0 | 1 | 0 | X | 1 |
| 1 | 0 | 1 | 1 | X |
| 1 | 1 | 0 | X | 0 |

K-maps



- A combinational circuit consists of logic gates whose outputs at any time are determined from only the present combination of inputs.
- A combinational circuit performs an operation that can be specified logically by a set of Boolean functions.
- It consists of an interconnection of logic gates. Combinational logic gates react to the values of the signals at their inputs and produce the value of the output signal, transforming binary information from the given input data to a required output data.
- A block diagram of a combinational circuit is shown in the below figure.
- The $n$ input binary variables come from an external source; the $m$ output variables are produced by the internal combinational logic circuit and go to an external destination.
- Each input and output variable exists physically as an analog signal whose values are interpreted to be a binary signal that represents logic 1 and logic 0 .



## BINARY ADDER-SUBTRACTOR:-

- Digital computers perform a variety of information-processing tasks. Among the functions encountered are the various arithmetic operations.
- The most basic arithmetic operation is the addition of two binary digits. This simple addition consists of four possible elementary operations: $0+0=0,0+1=1,1+0=1$, and $1+1=10$.
- The first three operations produce a sum of one digit, but when both augend and addend bits are equal to 1 ; the binary sum consists of two digits. The higher significant bit of this result is called a carry.
- When the augend and addend numbers contain more significant digits, the carry obtained from the addition of two bits is added to the next higher order pair of significant bits.
- A combinational circuit that performs the addition of two bits is called a half adder.
- One that performs the addition of three bits (two significant bits and a previous carry) is a full adder. The names of the circuits stem from the fact that two half adders can be employed to implement a full adder.


## HALF ADDER:-

- This circuit needs two binary inputs and two binary outputs.
- The input variables designate the augend and addend bits; the output variables produce the sum and carry. Symbols x and y are assigned to the two inputs and S (for sum) and C (for carry) to the outputs.
- The truth table for the half adder is listed in the below table.
- The C output is 1 only when both inputs are 1 . The S output represents the least significant bit of the sum.
- The simplified Boolean functions for the two outputs can be obtained directly from the truth table.

| $x$ | $y$ | $D$ | $B$ |
| :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |
| 0 | 1 | 1 | 0 |
| 1 | 0 | 1 | 0 |
| 1 | 1 | 0 | 1 |
| Truth Table |  |  |  |

- The simplified sum-of-products expressions are

$$
\begin{gathered}
S=\text { x'y }+ \\
\text { xy’ } C= \\
\text { xy }
\end{gathered}
$$

- The logic diagram of the half adder implemented in sum of products is shown in the below figure. It can
be also implemented with an exclusive-OR and an AND gate.

(a) $S-x y^{\prime}+x^{\prime} y$
$C=x y$

(b) $S-x \oplus y$


## FULL ADDER:-

- A full adder is a combinational circuit that forms the arithmetic sum of three bits.
- It consists of three inputs and two outputs. Two of the input variables, denoted by x and y , represent the two significant bits to be added. The third input, z , represents the carry from the previous lower

| $\boldsymbol{x}$ | $\boldsymbol{y}$ | $\boldsymbol{z}$ | $\mathbf{C}$ | $\boldsymbol{S}$ |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 0 | 1 |
| 0 | 1 | 0 | 0 | 1 |
| 0 | 1 | 1 | 1 | 0 |
| 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | 1 | 1 | 0 |
| 1 | 1 | 0 | 1 | 0 |
| 1 | 1 | 1 | 1 | 1 |

## Truth Table

significant position.

- Two outputs are necessary because the arithmetic sum of three binary digits ranges in value from 0 to 3 , and binary representation of 2 or 3 needs two bits. The two outputs are designated by the symbols S for sum and C for carry.

(a) $S=x^{\prime} y^{\prime} z+x^{\prime} y z^{\prime}+x y^{\prime} z^{\prime}+x y z$

(b) $C=x y+x z+y z$

K-Map for full adder

- The binary variable $S$ gives the value of the least significant bit of the sum. The binary variable C gives the output carry formed by adding the input carry and the bits of the words.
- The eight rows under the input variables designate all possible combinations of the three variables. The output variables are determined from the arithmetic sum of the input bits. When all input bits are 0 , the output is 0 .
- The $S$ output is equal to 1 when only one input is equal to 1 or when all three inputs are equal to 1 . The C output has a carry of 1 if two or three inputs are equal to 1 .
- The simplified expressions are

$$
S=x^{\prime} y^{\prime} z+x x^{\prime} y z^{\prime}+x y^{\prime} z^{\prime}+x y z
$$

$$
C=x y+x z+y z
$$

- The logic diagram for the full adder implemented in sum-of-products form is shown in figure.

- It can also be implemented with two half adders and one OR gate as shown in the figure.



## Implementation of Full Adder using Two Half Adders and an OR gate

- A full adder is a combinational circuit that forms the arithmetic sum of three
bits. BINARY ADDER:-
- A binary adder is a digital circuit that produces the arithmetic sum of two binary numbers.
- It can be constructed with full adders connected in cascade, with the output carry from each full adder connected to the input carry of the next full adder in the chain.
- Addition of n -bit numbers requires a chain of n full adders or a chain of one-half adder and $\mathrm{n}-1$ full adders. In the former case, the input carry to the least significant position is fixed at 0 .
- The interconnection of four full-adder (FA) circuits to provide a four-bit binary ripple carry adder is shown in the figure.
- The augend bits of A and the addend bits of B are designated by subscript numbers from right to left, with subscript 0 denoting the least significant bit.
- The carries are connected in a chain through the full adders. The input carry to the adder is C 0 , and it ripples through the full adders to the output carry C 4 . The S outputs generate the required sum bits.
- An n -bit adder requires n full adders, with each output carry connected to the input carry of the next higher order full adder.
- Consider the two binary numbers $A=1011$ and $B=0011$. Their sum $S=1110$ is formed with the fourbit adder as follows:

| Subscript i: | $\mathbf{3}$ | $\mathbf{2}$ | $\mathbf{1}$ | $\mathbf{0}$ |  |
| :--- | :---: | :---: | :---: | :---: | :---: |
| Input carry | 0 | 1 | 1 | 0 | $C_{i}$ |
| Augend | 1 | 0 | 1 | 1 | $A_{i}$ |
| Addend | 0 | 0 | 1 | 1 | $B_{i}$ |
| Sum | 1 | 1 | 1 | 0 | $S_{i}$ |
| Output carry | 0 | 0 | 1 | 1 | $C_{i+1}$ |

- The bits are added with full adders, starting from the least significant position (subscript 0 ), to form the sum bit and carry bit. The input carry $\mathrm{C}_{0}$ in the least significant position must be 0 .
- The value of $\mathrm{C}_{\mathrm{i}+1}$ in a given significant position is the output carry of the full adder. This value is transferred into the input carry of the full adder that adds the bits one higher significant position to the left.
- The sum bits are thus generated starting from the rightmost position and are available as soon as the corresponding previous carry bit is generated. All the carries must be generated for the correct sum bits to appear at the outputs.


Four Bit Binary Adder

## HALF SUBTRACTOR:-

- This circuit needs two binary inputs and two binary outputs.
- Symbols x and y are assigned to the two inputs and D (for difference) and B (for borrow) to the outputs.
- The truth table for the half subtractor is listed in the below table.

| $\boldsymbol{x}$ | $\boldsymbol{y}$ | $D$ | $B$ |  |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 |  |
| 0 | 1 | 1 | 1 |  |
| 1 | 0 | 1 | 0 |  |
| 1 | 1 | 0 | 0 |  |
| Truth Table |  |  |  |  |

- The B output is 1 only when the inputs are 0 and 1 . The D output represents the least significant bit of the subtraction.
- The subtraction operation is done by using the following rules as

$$
\begin{aligned}
& 0-0=0 ; \\
& 0-1=1 \text { with borrow } 1 ; \\
& 1-0=1 ; \\
& 1-1=0 .
\end{aligned}
$$

- The simplified Boolean functions for the two outputs can be obtained directly from the truth table. The simplified sum-of-products expressions are

$$
D=x^{\prime} y+x y^{\prime} \text { and } B=x \prime y
$$



- The logic diagram of the half adder implemented in sum of products is shown in the figure. It can be also implemented with an exclusive-OR and an AND gate with one inverted input.


## FULL SUBTRACTOR:-

- A full subtractor is a combinational circuit that forms the arithmetic subtraction operation of three bits.
- It consists of three inputs and two outputs. Two of the input variables, denoted by x and y , represent the two significant bits to be subtracted. The third input, z , is subtracted from the result of the first

| $x$ | $y$ | $z$ | $D$ | $B$ |
| :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 1 | 1 | 1 |
| 0 | 1 | 0 | 1 | 1 |
| 0 | 1 | 1 | 0 | 1 |
| 1 | 0 | 0 | 1 | 0 |
| 1 | 0 | 1 | 0 | 0 |
| 1 | 1 | 0 | 0 | 0 |
| 1 | 1 | 1 | 1 | 1 |

Truth Table
subtraction.

- Two outputs are necessary because the arithmetic subtraction of three binary digits ranges in value from 0 to 3 , and binary representation of 2 or 3 needs two bits. The two outputs are designated by the symbols D for difference and B for borrow.
- The binary variable D gives the value of the least significant bit of the difference. The binary variable B gives the output borrow formed during the subtraction process.



## K-Map for full Subtractor

- The eight rows under the input variables designate all possible combinations of the three variables. The output variables are determined from the arithmetic subtraction of the input bits.
- The difference D becomes 1 when any one of the input is 1 or all three inputs are equal to 1 and the borrow $B$ is 1 when the input combination is $\left(\begin{array}{lll}0 & 0 & 1\end{array}\right)$ or $\left(\begin{array}{lll}0 & 1 & 0\end{array}\right)$ or $\left(\begin{array}{lll}0 & 1 & 1\end{array}\right)$ or ( $\left.\begin{array}{lll}1 & 1 & 1\end{array}\right)$.
- The simplified expressions are

$$
\begin{aligned}
& D= x^{\prime} y^{\prime} z+ \\
& x y z x^{\prime} y z^{\prime}+x y^{\prime} z^{\prime}+ \\
& \text { x'z }+ \\
& \text { x'y }
\end{aligned}
$$

- The logic diagram for the full adder implemented in sum-of-products form is shown in figure.



## MAGNITUDE COMPARATOR:-

- A magnitude comparator is a combinational circuit that compares two numbers A and B and determines their relative magnitudes.
- The following description is about a 2-bit magnitude comparator circuit.
- The outcome of the comparison is specified by three binary variables that indicate whether $\mathrm{A}<\mathrm{B}, \mathrm{A}=$ B , or $\mathrm{A}>\mathrm{B}$.
- Consider two numbers, A and B, with two digits each. Now writing the coefficients of the numbers in descending order of significance:

$$
\begin{gathered}
\mathrm{A}=\mathrm{A}_{1} \\
\mathrm{~A}_{0} \mathrm{~B}= \\
\mathrm{B}_{1} \mathrm{~B}_{0}
\end{gathered}
$$

- The two numbers are equal if all pairs of significant digits are equal i.e. if and only if $\mathrm{A} 1=\mathrm{B} 1$, and $\mathrm{A} 0=$ B0.
- When the numbers are binary, the digits are either 1 or 0 , and the equality of each pair of bits can be expressed logically with an exclusive-NOR function as

$$
\mathrm{x} 1=\mathrm{A}_{1} \mathrm{~B}_{1}+\mathrm{A}_{1}^{\prime} \mathrm{B}_{1}^{\prime}
$$

$$
\text { And } \mathrm{x} 0=\mathrm{A}_{0} \mathrm{~B}_{0}+\mathrm{A}_{0}{ }^{\prime} \mathrm{B}_{0}{ }^{\prime}
$$

- The equality of the two numbers A and B is displayed in a combinational circuit by an output binary variable that we designate by the symbol $(\mathrm{A}=\mathrm{B})$.
- This binary variable is equal to 1 if the input numbers, A and B , are equal, and is equal to 0 otherwise.
- For equality to exist, all xi variables must be equal to 1 , a condition that dictates an AND operation of all variables:

$$
(\mathrm{A}=\mathrm{B})=\mathrm{x}_{1} \mathrm{x}_{0}
$$

- The binary variable $(\mathrm{A}=\mathrm{B})$ is equal to 1 only if all pairs of digits of the two numbers are equal.
- To determine whether $A$ is greater or less than $B$, we inspect the relative magnitudes of pairs of significant digits, starting from the most significant position. If the two digits of a pair are equal, we compare the next lower significant pair of digits. If the corresponding digit of A is 1 and that of B is 0 , we conclude that $A>B$. If the corresponding digit of $A$ is 0 and that of $B$ is 1 , we have $A<B$. The sequential comparison can be expressed logically by the two Boolean functions

$$
\begin{gathered}
(\mathrm{A}>\mathrm{B})= \\
\mathrm{A}_{1} \mathrm{~B}_{1}{ }^{\prime}+\mathrm{x}_{1} \mathrm{~A}_{0} \mathrm{~B}^{\prime}{ }_{0}(\mathrm{~A}<\mathrm{B}) \\
=\mathrm{A}_{1}{ }^{\prime} \mathrm{B}_{1}+\mathrm{x}_{1} \mathrm{~A}_{0}{ }^{\prime} \mathrm{B}_{0}{ }^{\prime}
\end{gathered}
$$

| Ai | A | B | Br | $\mathrm{A}>\mathrm{B}$ | $\mathrm{A}<\mathrm{B}$ | $\mathrm{A}=\mathrm{B}$ |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 0 | 0 | 0 | 1 | 0 |  | 0 |
| 0 | 0 | 1 | 0 | 0 | $i$ |  |
| 0 | 0 | 1 | 1 | 0 |  |  |
| 0 | 1 | 0 | 0 | 1 | 0 | 0 |
| 0 | 1 | 0 | 1 | 0 | 0 | 0 |
| 0 | 1 | 1 | 0 | 0 |  | 0 |
| 0 | 1 | 1 | 1 | 0 | $i$ |  |
| 1 | 0 | 0 | 0 |  | 0 | 0 |
| 1 | 0 | 0 | 1 |  | 0 | 0 |
| 1 | 0 | 1 | 0 | 0 | 0 | i |
| 1 | 0 | 1 | 1 | 0 | i | 0 |
| 1 | 1 | 0 | 0 |  | 0 | 0 |
| 1 | 1 | 0 | 1 | i | 0 | 0 |
| 1 | 1 | 1 | 0 | i | 0 | 0 |
| 1 | 1 | 1 | 1 | 0 | 0 | i |

Truth Table


Logic Diagram of 2-bit Magnitude Comparator


| $E$ | $A$ | $B$ | $D_{0}$ | $D_{1}$ | $D_{2}$ | $D_{3}$ |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 1 | $X$ | $X$ | 1 | 1 | 1 | 1 |
| 0 | 0 | 0 | 0 | 1 | 1 | 1 |
| 0 | 0 | 1 | 1 | 0 | 1 | 1 |
| 0 | 1 | 0 | 1 | 1 | 0 | 1 |
| 0 | 1 | 1 | 1 | 1 | 1 | 1 |

- A decoder is a combinational circuit that converts binary information from n input lines to a maximum of $2^{\mathrm{n}}$ unique output lines.
- If the n -bit coded information has unused combinations, the decoder may have fewer than 2 n outputs.
- The decoders presented here are called n -to- m -line decoders, where $\mathrm{m} . . .2 \mathrm{n}$.
- Their purpose is to generate the 2 n (or fewer) minterms of n input variables.
- Each combination of inputs will assert a unique output. The name decoder is also used in conjunction with other code converters, such as a BCD-to-seven-segment decoder.
- Consider the three-to-eight-line decoder circuit of three inputs are decoded into eight outputs, each representing one of the minterms of the three input variables.
- The three inverters provide the complement of the inputs, and each one of the eight AND gates generates one of the minterms.
- The input variables represent a binary number, and the outputs represent the eight digits of a number in the octal number system.
- However, a three-to-eight-line decoder can be used for decoding any three-bit code to provide eight outputs, one for each element of the code.
- A two-to-four-line decoder with an enable input constructed with NAND gates is shown in Fig.
- The circuit operates with complemented outputs and a complement enable input. The decoder is enabled when $E$ is equal to 0 (i.e., active-low enable). As indicated by the truth table, only one output can be equal to 0 at any given time; all other outputs are equal to 1 .
- The output whose value is equal to 0 represents the minterm selected by inputs A and B.
- The circuit is disabled when $E$ is equal to 1 , regardless of the values of the other two inputs.
- When the circuit is disabled, none of the outputs are equal to 0 and none of the minterms are selected.
- In general, a decoder may operate with complemented or un-complemented outputs.
- The enable input may be activated with a 0 or with a 1 signal.
- Some decoders have two or more enable inputs that must satisfy a given logic condition in order to enable the circuit.
- A decoder with enable input can function as a demultiplexer- a circuit that receives information from a single line and directs it to one of 2 n possible output lines.
- The selection of a specific output is controlled by the bit combination of $n$ selection lines.
- The decoder of Fig. can function as a one-to-four-line demultiplexer when E is taken as a data input line and $A$ and $B$ are taken as the selection inputs.
- The single input variable E has a path to all four outputs, but the input information is directed to only one of the output lines, as specified by the binary combination of the two selection lines A and B.
- This feature can be verified from the truth table of the circuit.
- For example, if the selection lines $\mathrm{AB}=10$, output $\mathrm{D}_{2}$ will be the same as the input value E , while all other outputs are maintained at 1 .
- Since decoder and demultiplexer operations are obtained from the same circuit, a decoder with an enable input is referred to as a decoder - demultiplexer.
- A application of this decoder is binary-to-octal conversion.


## ENCODER:-

- An encoder is a digital circuit that performs the inverse operation of a decoder.
- An encoder has 2 n (or fewer) input lines and n output lines.
- The output lines, as an aggregate, generate the binary code corresponding to the input value.

| Inputs |  |  |  |  |  |  |  | Outputs |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\mathrm{D}_{0}$ | $D_{1}$ | $D_{2}$ | $D_{3}$ | $\mathrm{D}_{4}$ | $D_{5}$ | $D_{6}$ | $D_{7}$ | $x$ | $\boldsymbol{r}$ | $z$ |
| 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 |
| 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 |
| 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 |
| 0 | 0 | 0 | 1 | 0 | 0 | 0 | 0 | 0 | 1 | 1 |
| 0 | 0 | 0 | 0 | 1 | 0 | 0 | 0 | 1 | 0 | 0 |
| 0 | 0 | 0 | 0 | 0 | 1 | 0 | 0 | 1 | 0 | 1 |
| 0 | 0 | 0 | 0 | 0 | 0 | 1 | 0 | 1 | 1 | 0 |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 |

- The above Encoder has eight inputs (one for each of the octal digits) and three outputs that generate the corresponding binary number.
- It is assumed that only one input has a value of 1 at any given time.
- The encoder can be implemented with OR gates whose inputs are determined directly from the truth table.
- Output z is equal to 1 when the input octal digit is $1,3,5$, or 7 .
- Output y is 1 for octal digits $2,3,6$, or 7 , and output x is 1 for digits $4,5,6$, or 7 .
- These conditions can be expressed by the following Boolean output functions:

$$
\begin{aligned}
& \mathrm{z}=\mathrm{D}_{1}+\mathrm{D}_{3}+\mathrm{D}_{5}+ \\
& \mathrm{D}_{7} \mathrm{y}=\mathrm{D}_{2}+\mathrm{D}_{3}+\mathrm{D}_{6} \\
& +\mathrm{D}_{7} \mathrm{x}=\mathrm{D}_{4}+\mathrm{D}_{5}+ \\
& \mathrm{D}_{6}+\mathrm{D}_{7}
\end{aligned}
$$

- The encoder can be implemented with three OR gates.
- The encoder defined above has the limitation that only one input can be active at any giventime.
- If two inputs are active simultaneously, the output produces an undefined combination.
- To resolve this ambiguity, encoder circuits must establish an input priority to ensure that only one input is encoded which is done in the Priority Encoder .


## PRIORITY ENCODER:-

- A priority encoder is an encoder circuit that includes the priority function.
- The operation of the priority encoder is such that if two or more inputs are equal to 1 at the same time, the input having the highest priority will take precedence.

| Inputs |  |  |  |  | Outputs |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| $\boldsymbol{D}_{\mathbf{0}}$ | $\boldsymbol{D}_{\mathbf{1}}$ | $\boldsymbol{D}_{\mathbf{2}}$ | $\boldsymbol{D}_{\mathbf{3}}$ |  | $\boldsymbol{x}$ | $\boldsymbol{y}$ | $\boldsymbol{V}$ |  |
| 0 | 0 | 0 | 0 |  | X | X | 0 |  |
| 1 | 0 | 0 | 0 |  | 0 | 0 | 1 |  |
| X | 1 | 0 | 0 |  | 0 | 1 | 1 |  |
| X | X | 1 | 0 |  | 1 | 0 | 1 |  |
| X | X | X | $\mathbf{1}$ |  | 1 | 1 | 1 |  |

- In addition to the two outputs x and y , the circuit has a third output designated by V ; this is a valid bit indicator that is set to 1 when one or
more inputs are equal to 1 .
- If all inputs are 0 , there is no valid input and V is equal to 0 .
- The other two outputs are not inspected when $V$ equals 0 and are specified as don't-care conditions.
- Here X 's in output columns represent don't-care conditions, the X 's in the input columns are useful for representing a truth table in condensed form.

| Inputs |  |  |  |  | Outputs |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
|  | $\boldsymbol{D}_{\mathbf{0}}$ | $\boldsymbol{D}_{\mathbf{1}}$ | $\boldsymbol{D}_{\mathbf{2}}$ | $\boldsymbol{D}_{\mathbf{3}}$ |  | $\boldsymbol{x}$ | $\boldsymbol{y}$ |
| 0 | 0 | 0 | 0 |  | X | X | 0 |
| X | 0 | 0 | 0 |  | 0 | 0 | 1 |
| X | 1 | 0 | 0 |  | 0 | 1 | 1 |
| X | X | 1 | 0 |  | 1 | 0 | 1 |
| X | X | X | 1 |  | 1 | 1 | 1 |

- Higher the subscript number, the higher the priority of the input.
- Input D3 has the highest priority, so, regardless of the values of the other inputs, when this input is 1 , the output for xy is 11 (binary 3 ).
- If D2 $=1$, provided that $\mathrm{D} 3=0$, regardless of the values of the other two lower priority inputs the output is 10 .
- The output for D1 is generated only if higher priority inputs are 0 , and so on down the priority levels.

- The maps for simplifying outputs x and y are shown in above Fig.
- The minterms for the two functions are derived from its truth table.
- Although the table has only five rows, when each $X$ in a row is replaced first by 0 and then by 1 , we obtain all 16 possible input combinations.
- For example, the fourth row in the table, with inputs XX10, represents the four minterms 0010, 0110, 1010, and 1110. The simplified Boolean expressions for the priority encoder are obtained from the maps.
- The condition for output V is an OR function of all the input variables.
- The priority encoder is implemented according to the following Boolean

$$
\text { functions: } \mathrm{x}=\mathrm{D}_{2}+\mathrm{D}_{3}
$$

$$
\begin{aligned}
\mathrm{y} & =\mathrm{D}_{3}+\mathrm{D}_{1} \mathrm{D}_{2}^{\prime} \\
\mathrm{V} & =\mathrm{D}_{0}+\mathrm{D}_{1}+\mathrm{D}_{2}+\mathrm{D}_{3}
\end{aligned}
$$



## MULTIPLEXER:-

- A multiplexer is a combinational circuit that selects binary information from one of many input lines and directs it to a single output line.
- The selection of a particular input line is controlled by a set of selection lines.
- Normally, there are $2^{\mathrm{n}}$ input lines and n selection lines whose bit combinations determine which input is selected.
- A four-to-one-line multiplexer is shown in the below figure. Each of the four inputs, $\mathrm{I}_{0}$ through $\mathrm{I}_{3}$, is applied to one input of an AND gate.
- Selection lines $S_{1}$ and $S_{0}$ are decoded to select a particular AND gate. The outputs of the AND gates are applied to a single OR gate that provides the one-line output.
- The function table lists the input that is passed to the output for each combination of the binary selection values.
- To demonstrate the operation of the circuit, consider the case when

$$
S_{1} S_{0}=10
$$

- The AND gate associated with input $\mathrm{I}_{2}$ has two of its inputs equal to 1 and the third input connected to $\mathrm{I}_{2}$.
- The other three AND gates have at least one input equal to 0 , which makes their outputs equal to 0 . The output of the OR gate is now equal to the value of $\mathrm{I}_{2}$, providing a path from the selected input to the output.
- A multiplexer is also called a data selector, since it selects one of many inputs and steers the binary information to the output line.

(b) Multiplexer implementation



Truth table

## DEMULTIPLEXER:-

- The data distributor, known more commonly as a Demultiplexer or "Demux" for short, is the exact opposite of the Multiplexer.
- The demultiplexer takes one single input data line and then switches it to any one of a number of individual output lines one at a time. The demultiplexer converts a serial data signal at the input to a parallel data at its output lines as shown below.
- The Boolean expression for this 1-to-4 demultiplexer above with outputs A to D and data select lines $\mathrm{a}, \mathrm{b}$ is given as:

$$
F=(a b)^{\prime} A+a^{\prime} b B+a b^{\prime} C+a b D
$$

- The function of the demultiplexer is to switch one common data input line to any one of the 4 output data lines A to D in our example above. As with the multiplexer the individual solid state switches are selected by the binary input address code on the output select pins "a" and "b" as shown.



Logic Diagram

- Unlike multiplexers which convert data from a single data line to multiple lines and demultiplexers which convert multiple lines to a single data line, there are devices available which convert data to and from multiple lines and in the next tutorial about combinational logic devices.
- Standard demultiplexer IC packages available are the TTL 74LS138 1 to 8-output demultiplexer, the TTL 74LS139 Dual 1-to-4 output demultiplexer or the CMOS CD4514 1-to-16 output demultiplexer.

| Output Select |  | Data output <br> Selected |
| :---: | :---: | :---: |
| b | a |  |
| 0 | 0 | A |
| 0 | 1 | B |
| 1 | 0 | C |
| 1 | 1 | D |
| Truth Table |  |  |
|  |  |  |

## COUNTER

- A counter is a device which stores (and sometimes displays) the number of times a particular event or process has occurred. In electronics, counters can be implemented quite easily using register-type circuits.
- There are different types of counters, viz.
- Asynchronous (ripple) counter
- Synchronous counter
- Decade counter
- Up/down counter
- Ring counter
- Johnson counter
- Cascaded counter
- Modulus counter.


## Synchronous counter

A 4-bit synchronous counter using JK flip-flops is shown in the figure.
In synchronous counters, the clock inputs of all the flip-flops are connected together and are triggered by the input pulses. Thus, all the flip-flops change state simultaneously (in parallel).


The circuit below is a 4-bit synchronous counter.
The J and K inputs of FF0 are connected to HIGH. FF1 has its J and K inputs connected to the output of FF0, and the J and K inputs of FF2 are connected to the output of an AND gate that is fed by the outputs of FF0 and FF1.
A simple way of implementing the logic for each bit of an ascending counter (which is what is depicted in the image to the right) is for each bit to toggle when all of the less significant bits are at a logic high state.
For example, bit 1 toggles when bit 0 is logic high; bit 2 toggles when both bit 1 and bit 0 are logic high; bit 3 toggles when bit 2, bit 1 and bit 0 are all high; and so on.

Synchronous counters can also be implemented with hardware finite state machines, which are more complex but allow for smoother, more stable transitions.

## Asynchronous Counter

An asynchronous (ripple) counter is a single d-type flip-flop, with its J (data) input fed from its own inverted output.
This circuit can store one bit, and hence can count from zero to one before it overflows (starts over from $0)$.


This counter will increment once for every clock cycle and takes two clock cycles to overflow, so every cycle it will alternate between a transition from 0 to 1 and a transition from 1 to 0 .
This creates a new clock with a $50 \%$ duty cycle at exactly half the frequency of the input clock.
If this output is then used as the clock signal for a similarly arranged D flip-flop, remembering to invert the output to the input, one will get another 1 bit counter that counts half as fast. These together yield a two-bit counter.
Additional flip-flops can be added, by always inverting the output to its own input, and using the output from the previous flip-flop as the clock signal. The result is called a ripple counter, which can count to $2^{\mathrm{n}}$
-1 , where n is the number of bits (flip-flop stages) in the counter.
Ripple counters suffer from unstable outputs as the overflows "ripple" from stage to stage, but they find application as dividers for clock signals.

## Modulus Counter

A modulus counter is that which produces an output pulse after a certain number of input pulses is applied.
In modulus counter the total count possible is based on the number of stages, i.e., digit positions.

Modulus counters are used in digital computers.
A binary modulo-8 counter with three flip-flops, i.e., three stages, will produce an output pulse, i.e., display an output one-digit, after eight input pulses have been counted, i.e., entered or applied. This assumes that the counter started in the zero-condition.

## Asynchronous Decade Counter


$\square$ A decade counter can count from BCD " 0 " to BCD " 9 ".
$\square$ A decade counter requires resetting to zero when the output count reaches the decimal value of 10 , ie. when DCBA $=1010$ and this condition is fed back to the reset input.
$\square$ A counter with a count sequence from binary " 0000 " ( $\mathrm{BCD}=$ " 0 ") through to " 1001 " ( $\mathrm{BCD}=$ " 9 ") is generally referred to as a BCD binary-coded-decimal counter because its ten state sequence is that of a BCD code but binary decade counters are more common.
$\square$ This type of asynchronous counter counts upwards on each leading edge of the input clock signal starting from 0000 until it reaches an output 1001 (decimal 9).
$\square$ Both outputs $Q_{A}$ and $Q_{D}$ are now equal to logic " 1 " and the output from the NAND gate changes state from logic " 1 " to a logic " 0 " level and whose output is also connected to the CLEAR ( CLR ) inputs of all the J-K Flip-flops.
$\square$ This signal causes all of the Q outputs to be reset back to binary 0000 on the count of 10 . Once QA and QD are both equal to logic " 0 " the output of the NAND gate returns back to a logic level " 1 " and the counter restarts again from 0000 . We now have a decade or Modulo- 10 counter.

## Decade Counter Truth Table

| Clock <br> Count | Output bit Pattern |  |  | Decimal |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
|  | QD | QC | QB | QA | Value |

## Up/Down Counter

In a synchronous up-down binary counter the flip-flop in the lowest-order position is complemented with every pulse.
$\square$ A flip-flop in any other position is complemented with a pulse, provided all the lower-order pulse equal to 0 .
$\square \mathrm{Up} /$ Down counter is used to control the direction of the counter through a certain sequence.


From the sequence table we can observe that:

- For both the UP and DOWN sequences, $\mathrm{Q}_{0}$ toggles on each clock pulse.
- For the UP sequence, $\mathrm{Q}_{1}$ changes state on the next clock pulse when $\mathrm{Q}_{0}=1$.
- For the DOWN sequence, $\mathrm{Q}_{1}$ changes state on the next clock pulse when $\mathrm{Q}_{0}=0$.
- For the UP sequence, $\mathrm{Q}_{2}$ changes state on the next clock pulse when $\mathrm{Q}_{0}=\mathrm{Q}_{1}=1$.
- For the DOWN sequence, $\mathrm{Q}_{2}$ changes state on the next clock pulse when $\mathrm{Q}_{0}=\mathrm{Q}_{1}=0$.

$\square$ These characteristics are implemented with the AND, OR \& NOT logic connected as shown in the logic diagram above.


## REGISTERS

## INTRODUCTION:-

The sequential circuits known as register are very important logical block in most of the digital systems.
$\square$ Registers are used for storage and transfer of binary information in a digital system.
$\square$ A register is mostly used for the purpose of storing and shifting binary data entered into it from an external source and has no characteristics internal sequence of states.
$\square$ The storage capacity of a register is defined as the number of bits of digital data, it can store or retain.
$\square$ These registers are normally used for temporary storage of data.

## BUFFER REGISTER:-

These are the simplest registers and are used for simply storing a binary word.
$\square$ These may be controlled by Controlled Buffer Register.
$\square$ D flip - flops are used for constructing a buffer register or other flip- flop can be used.
$\square$ The figure shown below is a 4- bit buffer register.

] When the clock pulse is applied, the output word becomes the same as the word applied at the input terminals, i.e. the input word is loaded into the register by the application of clock pulse.
When the positive clock edge arrives, the stored word becomes:

$$
\begin{aligned}
& \text { Q4 Q3 Q2 Q1 }=\text { X4 X3 X2 X1 } \\
& \text { or } \quad \mathrm{Q}=\mathrm{X} .
\end{aligned}
$$

This circuit is too primitive to be of any use.

## CONTROLLED BUFFER REGISTER:-

$\square$ The figure shows a controlled buffer register.

$\mathrm{Q}=0000$.
bvinell LLN is min, the rEgister is lEaky lvi action

LOAD is control input.
When LOAD is HIGH, the data bits X can reach the D inputs of FFs.
$\square$ At the positive going edge of the next clock pulse, the register is loaded, i.e.

$$
\begin{aligned}
& \text { Q4 Q3 Q2 Q1 }=\text { X4 X3 X2 X1 } \\
& \text { or } \quad \mathrm{Q}=\mathrm{X} .
\end{aligned}
$$

When LOAD is LOW, the X bits cannot reach the FFs. At the same time the inverted signal LOAD is HIGH. This forces each flip-flop output to feedback to its data input.
Therefore data is circulated or retained as each clock pulse arrives.
$\square$ In other words the content register remains unchanged in spite of the clock pulses.
Longer buffer registers can built by adding more FFs.

## CONTROLLED BUFFER REGISTER:-

$\square$ A number of FFs connected together such that data may be shifted into and shifted out of them is called a shift register.
Data may be shifted into or out of the register either in serial form or in parallel form.
$\square$ There are four basic types of shift registers

1. Serial in, serial out
2. Serial in, parallel out
3. Parallel in, serial out
4. Parallel in, parallel out

## SERIAL IN, SERIAL OUT SHIFT REGISTER:-

This type of shift register accepts data serially, i.e., one bit at a time and also outputs data serially.
$\square$ The logic diagram of a four bit serial in, serial out shift register is shown in below figure:
In 4 stages i.e. with 4 FFs , the register can store upto 4 bits of data.
$\square$ Serial data is applied at the D input of the first FF. The Q output of the first FF is connected to the D input of the second FF, the output of the second FF is connected to the D input of the third FF and the Q output of the third FF is connected to the D input of the fourth FF. The data is outputted from the Q terminal of the last FF.
$\square$ When a serial data is transferred to a register, each new bit is clocked into the first FF at the positive going edge of each clock pulse.
$\square$ The bit that is previously stored by the first FF is transferred to the second FF.
$\square$ The bit that is stored by the second FF is transferred to the third FF, and so on.
$\square$ The bit that was stored by the last FF is shifted out.
$\square$ A shift register can also be constructed using J-K FFs or S-R FFs as shown in the figure below.


4-bit serial-in, serial-out, shift-right, shift register.


## SERIAL IN, PARALLEL OUT SHIFT REGISTER:-

In this type of register, the data bits are entered into the register serially, but the data stored in the register serially, but the stored in the register is shifted out in the parallel form.When the data bits are stored once, each bits appears on its respective output line and all bits are available simultaneously, rather than bit - by - bit basis as in the serial output.
$\square$ The serial in, parallel out shift register can be used as a serial in, serial out shift register if the output is taken from the Q terminal of the last FF.
$\square$ The logic diagram and logic symbol of a 4 bit serial in, parallel out shift register is given below.


## A 4- bit serial in, parallel out shift register

## PARALLEL IN, SERIAL OUT SHIFT REGISTER:-

For parallel in, serial out shift register the data bits are entered simultaneously into their respective stages on parallel lines, rather than on bit by bit basis on one line as with serial data inputs, but the data bits are transferred out of the register serially, i.e., on a bit by bit basis over a single line.
$\square$ The logic diagram and logic symbol of 4 bit parallel in, serial out shift register using D FFs is shown below.
$\square$ There are four data lines $\mathrm{A}, \mathrm{B}, \mathrm{C}$ and D through which the data is entered into the register in parallel form.
The signal Shift /LOAD allows

1. The data to be entered in parallel form into the register and
2. The data to be shifted out serially from terminal $\mathrm{Q}_{4}$.
$\square$ When Shift / $\overline{\mathrm{LOAD}}$ line is HIGH, gates G1, G2, and G3 are disabled, but gates G4, G5 and G6 are enabled allowing the data bits to shift right from one stage to next.
$\square$ When Shift /LOAD line is LOW, gates G4, G5 and G6 are disabled, whereas gates G1, G2 and G3 are enabled allowing the data input to appear at the D inputs of the respective FFs.
$\square$ When clock pulse is applied, these data bits are shifted to the Q output terminals of the FFs and therefore the data is inputted in one step.
$\square$ The OR gate allows either the normal shifting operation or the parallel data entry depending on which AND gates are enabled by the level on the Shift /LOAD input.

(a) Logic diagram

(b) Logic symbol

## A 4- bit parallel in, serial out shift register

## PARALLEL IN, PARALLEL OUT SHIFT REGISTER:-

$\square$ In a parallel in, parallel out shift register, the data entered into the register in parallel form and also the data taken out of the register in parallel form. Immediately following the simultaneous entry of all data bits appear on the parallel outputs.
$\square$ The figure shown below is a 4 bit parallel in parallel out shift register using D FFs.
$\square$ Data applied to the D input terminals of the FFs.
When a clock pulse is applied at the positive edge of that pulse, the D inputs are shifted into the Q outputs of the FFs.
$\square$ The register now stores the data.
$\square$ The stored data is available instantaneously for shifting out in parallel form.


## Logic diagram of a 4 - bit parallel in, parallel out shift register

## BIDIRECTIONAL SHIFT REGISTER:-

In bidirectional shift register is one in which the data bits can be shifted from left to right or from right to left.
$\square$ The figure shown below the logic diagram of a 4 bit serial in, serial out, bidirectional ( shift-left, shiftright) shift register.
$\square$ Right /Left is the mode signal. When Right /Left is a 1, the logic circuit works as a shift right shift register. When Right /Left is a 0 , the logic circuit works as a shift right shift register.
$\square$ The bidirectional is achieved by using the mode signal and two AND gates and one OR gate for each stage.
$\square$ A HIGH on the Right/Left control input enables the AND gates $G_{1}, G_{2}, G_{3}$ and $G_{4}$ and disables the AND gates $\mathrm{G}_{5}, \mathrm{G}_{6}, \mathrm{G}_{7}$ and $\mathrm{G}_{8}$ and the state of Q output of each FF is passed through the gate to the D input of the following FF. When clock pulse occurs, the data bits are effectively shifted one place to the right.
$\square$ A LOW Right/Left control input enables the AND gates $G_{5}, G_{6}, G_{7}$ and $G_{8}$ and disables the AND gates $\mathrm{G}_{1}, \mathrm{G}_{2}, \mathrm{G}_{3}$ and $\mathrm{G}_{4}$ and the Q output of each FF is passed to the D input of the preceding FF. When clock pulse occurs the data bits are then effectively shifted one place to the left.
$\square$ So, the circuit works as a bidirectional shift register.


Logic diagram of 4-bit bidirectional shift register

## UNIVERSAL SHIFT REGISTERS:-

The register which has both shifts and parallel load capabilities, it is referred as a universal shift register. So, universal shift register is a bidirectional register, whose input can be either in serial form or in parallel form and whose output also can be either in serial form or parallel form.
$\square$ The universal shift register can be realized using multiplexers.
$\square$ The figure shows the logic diagram of a 4 bit universal shift register that has all the capabilities of a general shift register.


Fig- (a) 4 bit universal shift register

It consists of four D flip- flops and four multiplexers.
$\square$ The four multiplexers have two common selection inputs $S_{1}$ and $S_{0}$.
Input 0 in each multiplexer is selected when $S_{1} S_{0}=00$, input 1 is selected when $S_{1} S_{0}=01$, and input 2 is selected when $S_{1} S_{0}=10$ and input 3 is selected when $\quad S_{1} S_{0}=11$.
$\square$ The selection inputs control the mode of operation of the register is according to the function entries shown in the table.
When $S_{1} S_{0}=00$ the present value of the register is applied to the $D$ inputs of flip-flops. This condition forms a path from the output of each FF into the input of the same FF.
The next clock edge transfers into each FF the binary value it held previously, and no change of state occurs.
When $S_{1} S_{0}=01$, terminal 1 of the multiplexer inputs have a path of the $D$ inputs of the flip- flops. This causes a shift right operation, with serial input transferred into $\mathrm{FF}_{4}$.
When $S_{1} S_{0}=10$ a shift left operation results with the other serial input going into the $\mathrm{FF}_{1}$.
Finally when $S_{1} S_{0}=11$, the binary information on the parallel input lines is transferred into the register simultaneously during the next clock edge.

## Functional table for the register of fig - a:

| Mode control |  |  |
| :--- | :---: | :---: |
| $\mathrm{S}_{1}$ | $\mathrm{~S}_{0}$ | Register operation |
| 0 | 0 | No change |
| 0 | 1 | Shift right |
| 1 | 0 | Shift left |
| 1 | 1 | Parallel load |

## APPLICATIONS OF SHIFT REGISTERS:-

## 1. Time delays:

- In digital systems, it is necessary to delay the transfer of data until the operation of the other data have been completed, or to synchronize the arrival of data at a subsystem where it is processed with other data.
- A shift register can be used to delay the arrival of serial data by a specific number of clock
pulses, since the number of stages corresponds to the number of clock pulses required to shift each bit completely through the register.
- The total time delay can be controlled by adjusting the clock frequency and by the number of stages in the register.
- In practice, the clock frequency is fixed and the total delay can be adjusted only by controlling the number of stages through which the data is passed.

2. Serial / Parallel data conversion:

- Transfer of data in parallel form is much faster than that in serial form.
- Similarly the processing of data is much faster when all the data bits are available simultaneously. Thus in digital systems in which speed is important so to operate on data parallel form is used.
- When large data is to be transmitted over long distances, transmitting data on parallel lines is costly and impracticable.
- It is convenient and economical to transmit data in serial form, since serial data transmission requires only one line.
- Shift registers are used for converting serial data to parallel form, so that a serial input can be processed by a parallel system and for converting parallel data to serial form, so that parallel data can be transmitted serially.
- A serial in, parallel out shift register can be used to perform serial-to parallel conversion, and a parallel in, serial out shift register can be used to perform parallel- to -serial conversion.
- A universal shift register can be used to perform both the serial- to - parallel and parallel-to- serial data conversion.
- A bidirectional shift register can be used to reverse the order of data.


## Microprocessor

The Microprocessor is a programmable device that takes in numbers, performs on them arithmetic or logical operations according to the program stored in memory and then produces other numbers as a result (Silicon chip which includes ALU, register circuits \& control circuits).
A Programmable Machine can be represented with 4 components:-

1. Microprocessor
2. Input
3. Memory
4. Output


## Internal Architecture of 8085 Microprocessor:



8085 is an 8 -bit microprocessor designed by Intel in 1977 using NMOS technology.
It has the following configuration -

- 8-bit data bus
- 16-bit address bus, which can address upto 64 KB
- A 16-bit program counter
- A 16-bit stack pointer
- Six 8-bit registers arranged in pairs: BC, DE, HL
- Requires +5 V supply to operate at 3.2 MHZ single phase clock

It is used in washing machines, microwave ovens, mobile phones, etc.

## 8085 Microprocessor - Functional Units

8085 consists of the following functional units -

## Accumulator

It is an 8-bit register used to perform arithmetic, logical, I/O \& LOAD/STORE operations. It is connected to internal data bus \& ALU.

## Arithmetic and logic unit

As the name suggests, it performs arithmetic and logical operations like Addition, Subtraction, AND, OR, etc. on 8-bit data.

## General purpose register

There are 6 general purpose registers in 8085 processor, i.e. B, C, D, E, H \& L. Each register can hold 8-bit data.
These registers can work in pair to hold 16-bit data and their pairing combination is like B-C, D-E \& H-L.
Program counter
It is a 16-bit register used to store the memory address location of the next instruction to be executed. Microprocessor increments the program whenever an instruction is being executed, so that the program counter points to the memory address of the next instruction that is going to be executed.

## Stack pointer

It is also a 16-bit register works like stack, which is always incremented/decremented by 2 during push \& pop operations.

## Temporary register

It is an 8-bit register, which holds the temporary data of arithmetic and logical operations.

## Flag register

It is an 8 -bit register having five 1 -bit flip-flops, which holds either 0 or 1 depending upon the result stored in the accumulator.

These are the set of 5 flip-flops -

- $\quad$ Sign (S)
- Zero (Z)
- Auxiliary Carry (AC)
- Parity (P)
- Carry (C)

Its bit position is shown in the following table -

| D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| S | Z |  | AC |  | P |  | CY |

## Instruction register and decoder

It is an 8-bit register. When an instruction is fetched from memory then it is stored in the Instruction register. Instruction decoder decodes the information present in the Instruction register.

Timing and control unit
It provides timing and control signal to the microprocessor to perform operations. Following are the timing and control signals, which control external and internal circuits -

- Control Signals: READY, RD', WR', ALE
- Status Signals: S0, S1, IO/M'
- DMA Signals: HOLD, HLDA
- RESET Signals: RESET IN, RESET OUT

Interrupt control
As the name suggests it controls the interrupts during a process. When a microprocessor is executing a main program and whenever an interrupt occurs, the microprocessor shifts the control from the main program to process the incoming request. After the request is completed, the control goes back to the main program.

There are 5 interrupt signals in 8085 microprocessor: INTR, RST 7.5, RST 6.5, RST 5.5, TRAP.

## Serial Input/output control

It controls the serial data communication by using these two instructions: SID (Serial input data) and SOD (Serial output data).

Address buffer and address-data buffer
The content stored in the stack pointer and program counter is loaded into the address buffer and address-data buffer to communicate with the CPU. The memory and I/O chips are connected to these buses; the CPU can exchange the desired data with the memory and I/O chips.

Address bus and data bus
Data bus carries the data to be stored. It is bidirectional, whereas address bus carries the location to where it should be stored and it is unidirectional. It is used to transfer the data \& Address I/O devices.

## The main features of 8085 microprocessor are:

- It is an 8 bit microprocessor.
- It is manufactured with N-MOS technology.
- It has 16-bit address bus and hence can address up to $216=65536$ bytes (64KB) memory locations through A0-A15
- The first 8 lines of address bus and 8 lines of data bus are multiplexed AD0 - AD7
- Data bus is a group of 8 lines D0 - D7
- It supports external interrupt request.
- A 16 bit program counter (PC)
- A 16 bit stack pointer (SP)
- Six 8-bit general purpose register arranged in pairs: BC, DE, HL.
- It requires a signal +5 V power supply and operates at 3.2 MHZ single phase clock.
- It is enclosed with 40 pins DIP (Dual in line package).

The following image depicts the pin diagram of 8085 Microprocessor -


The pins of a 8085 microprocessor can be classified into seven groups -

## Address bus

A15-A8, it carries the most significant 8-bits of memory/IO address.

## Data bus

AD7-AD0, it carries the least significant 8-bit address and data bus.

## Control and status signals

These signals are used to identify the nature of operation. There are 3 control signal and 3 status signals.
Three control signals are RD, WR \& ALE.

- RD - This signal indicates that the selected IO or memory device is to be read and is ready for accepting data available on the data bus.
- WR - This signal indicates that the data on the data bus is to be written into a selected memory or IO location.
- ALE - It is a positive going pulse generated when a new operation is started by the microprocessor. When the pulse goes high, it indicates address. When the pulse goes down it indicates data.
Three status signals are IO/M, S0 \& S1.


## IO/M

This signal is used to differentiate between IO and Memory operations, i.e. when it is high indicates IO operation and when it is low then it indicates memory operation.

## S1 \& S0

These signals are used to identify the type of current operation.

## Power supply

There are 2 power supply signals - VCC \& VSS. VCC indicates $+5 v$ power supply and VSS indicates ground signal.

## Clock signals

There are 3 clock signals, i.e. X1, X2, CLK OUT.

- $\mathbf{X 1}, \mathbf{X 2}$ - A crystal (RC, LC N/W) is connected at these two pins and is used to set frequency of the internal clock generator. This frequency is internally divided by 2.
- CLK OUT - This signal is used as the system clock for devices connected with the microprocessor.

Interrupts \& externally initiated signals
Interrupts are the signals generated by external devices to request the microprocessor to perform a task. There are 5 interrupt signals, i.e. TRAP, RST 7.5, RST 6.5, RST 5.5, and INTR. We will discuss interrupts in detail in interrupts section.

- INTA - It is an interrupt acknowledgment signal.
- RESET IN - This signal is used to reset the microprocessor by setting the program counter to zero.
- RESET OUT - This signal is used to reset all the connected devices when the microprocessor is reset.
- READY - This signal indicates that the device is ready to send or receive data. If READY is low, then the CPU has to wait for READY to go high.
- HOLD - This signal indicates that another master is requesting the use of the address and data buses.
- HLDA (HOLD Acknowledge) - It indicates that the CPU has received the HOLD request and it will relinquish the bus in the next clock cycle. HLDA is set to low after the HOLD signal is removed.


## Serial I/O signals

There are 2 serial signals, i.e. SID and SOD and these signals are used for serial communication.

- SOD (Serial output data line) - The output SOD is set/reset as specified by the SIM instruction.
- SID (Serial input data line) - The data on this line is loaded into accumulator whenever a RIM instruction is executed.

Now let us discuss the addressing modes in 8085 Microprocessor.

## Addressing Modes in 8085

These are the instructions used to transfer the data from one register to another register, from the memory to the register, and from the register to the memory without any alteration in the content. Addressing modes in 8085 is classified into 5 groups -

Immediate addressing mode
In this mode, the 8/16-bit data is specified in the instruction itself as one of its operand. For example: MVI K, 20F: means 20 F is copied into register K .

## Register addressing mode

In this mode, the data is copied from one register to another. For example: MOV K, B: means data in register B is copied to register K.

## Direct addressing mode

In this mode, the data is directly copied from the given address to the register. For example: LDB 5000K: means the data at address 5000 K is copied to register B .

Indirect addressing mode
In this mode, the data is transferred from one register to another by using the address pointed by the register. For example: MOV K, B: means data is transferred from the memory address pointed by the register to the register K.
Implied addressing mode
This mode doesn't require any operand; the data is specified by the opcode itself. For example: CMP.

## Interrupts in 8085

Interrupts are the signals generated by the external devices to request the microprocessor to perform a task. There are 5 interrupt signals, i.e. TRAP, RST 7.5, RST 6.5, RST 5.5, and INTR.
Interrupt are classified into following groups based on their parameter -

- Vector interrupt - In this type of interrupt, the interrupt address is known to the processor. For example: RST7.5, RST6.5, RST5.5, TRAP.
- Non-Vector interrupt - In this type of interrupt, the interrupt address is not known to the processor so, the interrupt address needs to be sent externally by the device to perform interrupts. For example: INTR.
- Maskable interrupt - In this type of interrupt, we can disable the interrupt by writing some instructions into the program. For example: RST7.5, RST6.5, RST5.5.
- Non-Maskable interrupt - In this type of interrupt, we cannot disable the interrupt by writing some instructions into the program. For example: TRAP.
- Software interrupt - In this type of interrupt, the programmer has to add the instructions into the program to execute the interrupt. There are 8 software interrupts in 8085 , i.e. RST0, RST1, RST2, RST3, RST4, RST5, RST6, and RST7.
- Hardware interrupt - There are 5 interrupt pins in 8085 used as hardware interrupts, i.e. TRAP, RST7.5, RST6.5, RST5.5, INTA.

Note - NTA is not an interrupt, it is used by the microprocessor for sending acknowledgement. TRAP has the highest priority, then RST7.5 and so on.

## Interrupt Service Routine (ISR)

A small program or a routine that when executed, services the corresponding interrupting source is called an ISR.
TRAP
It is a non-maskable interrupt, having the highest priority among all interrupts. Bydefault, it is enabled until it gets acknowledged. In case of failure, it executes as ISR and sends the data to backup memory. This interrupt transfers the control to the location 0024 H .

## RST7.5

It is a maskable interrupt, having the second highest priority among all interrupts. When this interrupt is executed, the processor saves the content of the PC register into the stack and branches to 003 CH address.

## RST 6.5

It is a maskable interrupt, having the third highest priority among all interrupts. When this interrupt is executed, the processor saves the content of the PC register into the stack and branches to 0034 H address.

## RST 5.5

It is a maskable interrupt. When this interrupt is executed, the processor saves the content of the PC register into the stack and branches to 002 CH address.

INTR
It is a maskable interrupt, having the lowest priority among all interrupts. It can be disabled by resetting the microprocessor.

When INTR signal goes high, the following events can occur -

- The microprocessor checks the status of INTR signal during the execution of each instruction.
- When the INTR signal is high, then the microprocessor completes its current instruction and sends active low interrupt acknowledge signal.
- When instructions are received, then the microprocessor saves the address of the next instruction on stack and executes the received instruction.

Let us take a look at the programming of 8085 Microprocessor.
Instruction sets are instruction codes to perform some task. It is classified into five categories.
S.No.

## Instruction \& Description

## Control Instructions

1
Following is the table showing the list of Control instructions with their meanings.

## Logical Instructions

2 Following is the table showing the list of Logical instructions with their meanings.

Branching Instructions
3 Following is the table showing the list of Branching instructions with their meanings.

Arithmetic Instructions
4 Following is the table showing the list of Arithmetic instructions with their meanings.

## Data Transfer Instructions

5 Following is the table showing the list of Data-transfer instructions with their meanings.

## Programmable peripheral interface 8255

PPI 8255 is a general purpose programmable I/O device designed to interface the CPU with its outside world such as ADC, DAC, keyboard etc. We can program it according to the given condition. It can be used with almost any microprocessor.

It consists of three 8-bit bidirectional I/O ports i.e. PORT A, PORT B and PORT C. We can assign different ports as input or output functions


It consists of 40 pins and operates in +5 V regulated power supply. Port C is further divided into two 4 -bit ports i.e. port C lower and port C upper and port C can work in either BSR (bit set rest) mode or in mode 0 of input-output mode of 8255 . Port B can work in either mode 0 or in mode 1 of input-output mode. Port A can work either in mode 0 , mode 1 or mode 2 of input-output mode.

It has two control groups, control group A and control group B. Control group A consist of port A and port C upper. Control group B consists of port C lower and port B .
Depending upon the value if CS', A1 and A0 we can select different ports in different modes as input-output function or BSR. This is done by writing a suitable word in control register (control word D0-D7).

CS' A1 A0 Selection Address

| CS' | A1 | A0 | Selection | Address |
| :--- | :--- | :--- | :--- | :--- |
| 0 | 0 | 0 | PORT A | 80 H |
| 0 | 0 | 1 | PORT B | 81 H |
| 0 | 1 | 0 | PORT C | 82 H |
| 0 | 1 | 1 | Control Register | 83 H |
| 1 | X | X | No Selection | X |

## Pin diagram -

| PA3 $\leftrightarrow{ }^{1}$ |  | 40 | $\leftrightarrow$ PA4 |
| :---: | :---: | :---: | :---: |
| PA2 $\leftrightarrow 2$ |  | 39 | $\leftrightarrow$ PA5 |
| PA1 $\leftrightarrow 3$ |  | 38 | $\leftrightarrow$ PA6 |
| $\mathrm{PAO} \leftrightarrow 4$ |  | 37 | $\leftrightarrow$ PA7 |
| $\mathrm{RD}^{\prime} \rightarrow 5$ |  | 36 | $\leftarrow \mathrm{WR}^{\prime}$ |
| $\mathrm{CS} \rightarrow 6$ |  | 35 | $\leftarrow$ RESE |
| GND 47 |  | 34 | $\leftrightarrow$ D0 |
| VSS $\rightarrow 8$ |  | 33 | $\leftrightarrow$ D1 |
| A1 $\rightarrow 9$ |  | 32 | $\leftrightarrow$ D2 |
| A0 $\leftrightarrow 10$ | 8255 | 31 | $\leftrightarrow$ D3 |
| $\mathrm{PC7} \leftrightarrow 11$ |  | 30 | $\leftrightarrow$ D4 |
| PC6 $\leftrightarrow 12$ |  | 29 | $\leftrightarrow$ D5 |
| PC5 $\leftrightarrow 13$ |  | 28 | $\leftrightarrow$ D6 |
| $\mathrm{PC} 4 \longleftrightarrow 14$ |  | 27 | $\leftrightarrow$ D7 |
| $\mathrm{PCO} \leftrightarrow 15$ |  | 26 | $\leftarrow$ VCC |
| $\mathrm{PC} 2 \leftrightarrow 16$ |  | 25 | $\leftrightarrow \mathrm{PB} 7$ |
| PC3 $\leftrightarrow 17$ |  | 24 | $\leftrightarrow$ PB6 |
| $\mathrm{PBO} \leftrightarrow 18$ |  | 23 | $\leftrightarrow$ PB5 |
| PB1 $\longleftrightarrow 19$ |  | 22 | $\leftrightarrow$ PB4 |
| PB2 $\leftrightarrow 20$ |  | 21 | $\leftrightarrow \mathrm{PB} 3$ |

- PA0 - PA7 - Pins of port A
- PB0 - PB7 - Pins of port B
- PC0 - PC7 - Pins of port C
- D0 - D7 - Data pins for the transfer of data
- RESET - Reset input
- RD' - Read input
- WR' - Write input
- CS' - Chip select
- A1 and A0 - Address pins

Operating modes -

1. Bit set reset (BSR) mode -

If MSB of control word (D7) is 0, PPI works in BSR mode. In this mode only port C bits are used for set or

2. Input-Output mode -

CONTROL REGISTER INPUT-OUTPUT MODE

3. If MSB of control word (D7) is 1, PPI works in input-output mode. This is further divided into three modes:

- Mode $\mathbf{0}$-In this mode all the three ports (port A, B, C) can work as simple input function or simple output function. In this mode there is no interrupt handling capacity.
- Mode 1 - Handshake I/O mode or strobbed I/O mode. In this mode either port A or port B can work as simple input port or simple output port, and port C bits are used for handshake signals before actual data transmission. It has interrupt handling capacity and input and output are latched.
Example: A CPU wants to transfer data to a printer. In this case since speed of processor is very fast as compared to relatively slow printer, so before actual data transfer it will send handshake signals to the printer for synchronization of the
speed of the CPU and the peripherals.

- Mode 2 - Bi-directional data bus mode. In this mode only port A works, and port B can work either in mode 0 or mode 1.6 bits port C are used as handshake signals. It also has interrupt handling capacity.

Aim: Write 8085 assembly language program for addition of two 8 -bit numbers
Program

| Memory <br> address | Machine <br> Codes | Mnemonics | Comments |
| :---: | :---: | :---: | :--- |
| 8500 | 21 | LXI H, 8000 H | Address of first number in H-L <br> registerpair. |
| 8501 | 00 |  |  |
| 8502 | 80 |  |  |
| 8503 | 7 E | MOVA,M | Transfer first number in accumulator. |
| 8504 | 23 | INXH | Increment content of H-L register pair |
| 8505 | 66 | ADDM | Add first number and second number |
| 8506 | 32 | STA8003H | Store sum in 8003 H |
| 8507 | 03 |  |  |
| 8508 | 80 |  |  |
| 8509 | 76 | HLT | Halt |

Aim-Write 8085 an assembly language program for subtraction of two 8 -bit numbers.

## Program

| Memory <br> address | Machine <br> Codes | Mnemonics | Comments |
| :---: | :---: | :---: | :--- |
| 8500 | 21 | LXI H, 8000 H | Address of first number in H-L register <br> pair. |
| 8501 | 00 |  |  |
| 8502 | 80 |  | MOVA,M | Transfer first number in accumulator..

Aim: To write an assembly language for multiplying two 8 bit numbers by using 8085 micro processor kit.

## Program

| Memory address | Label | Mnemonics |  | Hex Code | Comments |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 8500 |  | MVI | A, 03 | 3E | $\mathrm{A}=00 \mathrm{H}$ |
| 8501 |  |  |  | 03 |  |
| 8502 |  | MOV | E,A | 5F | $\mathrm{E}=\mathrm{A}$. |
| 8503 |  | MVI | D, 00 | 16 | Get the first numberin |
| 8504 |  |  |  | 00 | DE register pair |
| 8505 |  | LDA | 8000 | 3A |  |
| 8506 |  |  |  | 00 | Store the content of |
| 8507 |  |  |  | 80 |  |
| 8508 |  | MOV | C,A | 4F | Initialize counter |
| 8509 |  | LXI | H, 0000 | 21 |  |
| 850A |  |  |  | 00 | Result $=0$ |
| 850B |  |  |  | 00 |  |
| 850C | BACK | DAD | $14^{\text {D }}$ | 19 | Result $=$ Result + first number |

Aim: To write an assembly language for multiplying two 8 bit numbers by using 8085 micro processor kit.

## Program

| Memory address | Label | Mnemonics |  | Hex Code | Comments |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 8500 |  | MVI | A, 03 | 3E | $\mathrm{A}=00 \mathrm{H}$ |
| 8501 |  |  |  | 03 |  |
| 8502 |  | MOV | E,A | 5F | $\mathrm{E}=\mathrm{A}$. |
| 8503 |  | MVI | D, 00 | 16 | Get the first numberin |
| 8504 |  |  |  | 00 | DE register pair |
| 8505 |  | LDA | 8000 | 3A |  |
| 8506 |  |  |  | 00 |  |
| 8507 |  |  |  | 80 |  |
| 8508 |  | MOV | C, A | 4F | Initialize counter |
| 8509 |  | LXI | H, 0000 | 21 |  |
| 850A |  |  |  | 00 | Result $=0$ |
| 850B |  |  |  | 00 |  |
| 850C | BACK | DAD | D | 19 | Result $=$ Result + first number |

Aim: To write an assembly language program for dividing two 8 bit numbers using microprocessor kit.

## Program

| Memoy address | Label | Mnemonics |  | Hex Code | Comments |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 8500 |  | MVI | C, 00 | 0E | Initialize Quotient as |
| 8501 |  |  |  | 00 | zero |
| 8502 |  | LDA | 8000 | 3A |  |
| 8503 |  |  |  | 00 |  |
| 8504 |  |  |  | 80 |  |
| 8505 |  | MOV | B,A | 47 | Copy the 1st data into register B |
| Memoy address | Label | Mn | onics | Hex Code | Comments |
| 8506 |  | LDA | 8001 | 3A | Get the second number |
| 8507 |  |  |  | 01 | in Accumulator |
| 8508 |  |  |  | 80 |  |
| 8509 |  | CMP | B | B8 | Compare the 2 values |
| 850A |  | JC | $\begin{gathered} \text { LOO } \\ \text { P1 } \end{gathered}$ | DA | Jump if dividend lesser |
| 850B |  |  |  | 12 | than divisor |
| 850C |  |  |  | 85 |  |
| 850D | LOOP2 | SUB | $\begin{gathered} \text { B } \\ 15 \end{gathered}$ | 90 | Subtract the 1st value by 2 ndvalue |


| 850E |  | INR | C | 0C | Increment Quotient |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 850F |  | JMP |  | C3 | Jump to Loop 1 till the value of dividend becomes zero |
| 8510 |  |  |  | 0D |  |
| 8511 |  |  |  | 85 |  |
| 8512 | LOOP1 | STA | 8002 | 32 | Store result |
| 8513 |  |  |  | 02 |  |
| 8514 |  |  |  | 80 |  |
| 8515 |  | MOV | A, C | 79 | Move the value of remainder accumulator |
| 8516 |  | STA | 8003 | 32 | Store theremaindervalue inaccumulator |
| 8517 |  |  |  | 03 |  |
| 8518 |  |  |  | 80 |  |
| 8519 |  | HLT |  |  | Stop execution |

Aim: Write 8085 assembly language program for one's complement of an 8 -bit numbers

## Program

| Memory Address | Hex Code | Mnemonics | Comments |
| :---: | :---: | :---: | :---: |
| 8500 | 21 | LXIH,8000H | Load address of number in $\mathrm{H}-\mathrm{L}$ register pair |
| 8501 | 00 |  |  |
| 8502 | 80 |  |  |
| 8503 | 7 E | MOVA,M | Move number into accumulator |
| 8504 | 3F | CMA | Complement accumulator |
| 8505 | 32 | STA8050H | Store the result in 8050 H |
| 8506 | 50 |  |  |
| 8507 | 80 |  |  |
| 8508 | 76 | HLT | Stop Execution |

Aim: Write 8085 assembly language program for two's complement of an 8 -bit numbers

## Program

| Memory Address | Hex Code | Mnemonics | Comments |
| :---: | :---: | :---: | :--- |
| 8500 | 21 | LXIH,8000H | Load address of number in H-L <br> register pair |
| 8501 | 00 |  |  |
| 8502 | 80 |  | Move number into accumulator |
| 8503 | 7 E | MOVA,M | Complement accumulator |
| 8504 | 3 F | CMA | Com |
| 8505 | C6 | ADI 01 | Add 01H with accumulator to <br> find two's complement of <br> number |
| 8506 | 01 | 01 | Store the result in 8050H |
| 8507 | 32 | STA8050H |  |
| 8508 | 50 |  | Stop Execution |
| 8509 | 80 | 16 |  |
| 850 A | 76 | HLT |  |

Aim: To find the largest element in an array of size ' $n$ ' using 8085 Microprocessor.

## Program

| Memory address | Label | Mnemonics | Hex Code | Comments |
| :---: | :---: | :---: | :---: | :---: |
| 8500 |  | LDA 8000 | 3A | Load the number of values |
| 8501 |  |  | 00 |  |
| 8502 |  |  | 80 |  |
| 8503 |  | MOV C,A | 79 | Initialize counter |
| 8504 |  | XRA A | AF | Clear Accumulator |
| 8505 |  | LXI H, 8001 | 21 | Set the pointer for array |
| 8506 |  |  | 01 |  |
| 8507 |  |  | 80 |  |
| 8508 | BACK | CMP M | BD | Is number> maximum |
| 8509 |  | JNC SKIP | D2 | No, jump to SKIP |
| 850A |  |  | 0D |  |
| 850B |  |  | 85 |  |
| 850C |  | MOV A,M | 7 E | replace maximum |
| 850D | SKIP | INX H | 23 | Increment pointer |
| 850 E |  | DCR C | 0D | Decrement counter by one |
| 850F |  | JNZ BACK | C2 | Go to next iteration |
| 8510 |  |  | 08 |  |
| 8511 |  |  | 85 |  |

Aim-Write a program to control the traffic light system using $8085 \& 8255 \mathrm{ppi}$.

## Program

| Memory <br> Address | Label | Machine <br> Code | Mnemonics | Operands | Comments |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 2000 |  | 3E 80 | MVI | A,80H | Init PA \&PB as output |
| 2002 |  | D3 03 | OUT | 03 H |  |
| 2004 |  | $3 E 11$ | MVI | A,11H | Stop all four ends |
| 2006 |  | D3 00 | OUT | 00 H |  |
| 2008 |  | D3 02 | OUT | 02H |  |
| 200A |  | CD 50 20 | CALL | DELAY1 |  |
| 200 D | LOOP | 3E 44 | MVI | A,44H | GO STR signal of <br> North \& South, <br> STOP signal of East <br> \&West |
| 200 F |  |  |  | OUT | 00 H |
| 2011 |  |  | CALL | DELAY1 |  |
| 2014 |  |  | OUT | A,22H | Alert signal for traffic |
| 2016 |  |  | CALL | 00H |  |
| 2018 |  |  | MVI 17 | A,99H | GO LEFT signal of <br> North \& South |
| 201 B |  |  |  |  |  |


| 201 D |  |  | OUT | 00 H |  |
| :--- | :--- | :--- | :--- | :--- | :--- |
| 201 F |  |  | CALL | DELAY1 |  |
| 2022 |  |  | MVI | A,22H | Alert signal for traffic |
| 2024 |  |  | OUT | 00 H |  |
| 2026 |  |  | CALL | DELAY2 |  |
| 2029 |  |  | MVI | A,11H |  <br> South |
| 202 B |  |  | OUT | 00 H |  |
| 202 D |  |  | OUT | A,44H | GO STR signal of East <br> \& West |
| 202 F |  |  | CALL | DELAY1 |  |
| 2031 |  |  | OUT | A,22H | Alert signal for traffic |
| 2034 |  |  | CALL | DELAY2 |  |
| 2036 |  |  |  |  |  |
| 2038 |  |  |  |  |  |


| Memory Address | Label | Machine Code | Mnemonics | Operands | Comments |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 203B |  |  | MVI | A,99H | GO Left signal of East \& West |
| 203D |  |  | OUT | 02H |  |
| 203F |  |  | CALL | DELAY1 |  |
| 2042 |  |  | MVI | A,22H | Alert signal for traffic |
| 2044 |  |  | OUT | 02H |  |
| 2046 |  |  | CALL | DELAY2 |  |
| 2049 |  |  | MVI | A, 11H | STOP signal of East \&West |
| 204B |  |  | OUT | 02H |  |
| 204D |  |  | JMP | LOOP | Jump to loop |
| 2050 |  | DELAY1: | MVI | B,25H | Delay of 10 sec . |
| 2052 |  | LP3: | MVI | C,0FFH |  |
| 2054 |  | LP2: | MVI | D, 0FFH |  |
| 2056 |  | LP1: | DCR | D |  |
| 2057 |  |  | JNZ | LP1 |  |
| 205A |  |  | DCR | C |  |
| 205B |  |  | JNZ | LP2 |  |
| 205 E |  |  | DCR | B |  |
| 205F |  |  | JNZ | LP3 |  |
| 2062 |  |  | RET |  |  |
| 2063 |  | DELAY2: | MVI | B,05H | Delay of 2 sec |
| 2065 |  | LP6: | MVI | C,0FFH |  |
| 2067 |  | LP5: | MVI | D,0FFH |  |
| 2069 |  | LP4: | DCR | D |  |
| 206A |  |  | JNZ | LP4 |  |
| 206D |  |  | DCR | C |  |
| 206E |  |  | JNZ | LP5 |  |
| 2071 |  |  | DCR | B |  |
| 2072 |  |  | JNZ | LP6 |  |
| 2075 |  |  | $\mathrm{RET}_{18}$ |  |  |

Aim: Program for Decimal Addition of Two 8-Bit Numbers and Sum is 16 Bit.

## Program

| Memory <br> Address | Machine Codes (Data) | Labels | Mnemonics | Operands | Comments |
| :---: | :---: | :---: | :---: | :---: | :---: |
| 8000 | 21 |  | LXI | H, 8501 H | Address of first number in H-L register pair. |
| 8001 | 01 |  |  |  | Lower byte data isstored in memory |
| 8002 | 85 |  |  |  | Higher byte data is stored in memory |
| 8003 | 0E |  | MVI | C,00H | Sum of msb's\& register value in $00 h$ |
| 8004 | 00 |  |  |  |  |
| 8005 | 7E |  | MOV | A,M | Transfer first number in accumulator. |
| 8006 | 23 |  | INX | H | Increment content of H-L register pair |
| 8007 | 86 |  | ADD | M | Add first number and second number |
| 8008 | 27 |  | DAA |  |  |
| 8009 | D2 |  | JNC | 800CH | Jump if no carry to 800Ch location |
| 800A | 0C |  |  |  | Lower byte data isstored in memory |
| 800B | 80 |  |  |  | Higher byte data is stored in memory |
| 800C | 0C |  | INR | C | Increment register C |
| 800D | 32 | AHEAD | STA | 8503H | Data of accumulatoris stored into 8503haddress |
| 800E | 03 |  |  |  | Lower byte data isstored in memory |
| 800F | 85 |  |  |  | Higher byte data is stored in memory |
| 8010 | 79 |  | MOV | A,C | MSB'S of sum in A |


| 8011 | 32 |  | STA | 8504 H | MSB'S of sum in A is <br> transferred to <br> 8504h |
| :---: | :---: | :---: | :---: | :---: | :--- |
| 8012 | 04 |  |  |  | location |
| isstored byte data |  |  |  |  |  |
| in memory |  |  |  |  |  |

Aim: Program for Decimal Subtraction of Two 8-Bit Numbers

## Program

| Memory <br> Address | Machine <br> Codes (Data) | Labels | Mnemonics | Operands | Comments |
| :---: | :---: | :---: | :---: | :---: | :--- |
| 8000 | 21 |  | LXI | H, 8501 H | Address of first number <br> in H-L register pair. |
| 8001 | 01 |  |  | Lower byte data isstored <br> in memory |  |
| 8002 | 85 |  |  | Higher byte data is stored <br> in memory |  |
| 8003 | 3 E |  | MVI | A,99H | Copy immediate data 99 <br> in A |
| 8004 | 99 |  |  | M | 9's complement |
| 8005 | 96 |  | SUB | M | Increment content of A <br> register |
| 8006 | $3 C$ |  | INR | A | Decrement content of H- <br> L register pair |
| 8007 | $2 B$ |  | DCX | Addition <br> complemented data and <br> the second number |  |
| 8008 | 86 |  | ADD | M | Decimal Accumulator <br> Adjust |
| 8009 | 27 |  | DAA |  | Data of accumulatoris <br> stored into 8503haddress |
| 800 A | 32 |  | RST.5 |  | 8503H |
| $800 B$ | 03 |  |  |  |  |
| $800 C$ | 85 |  |  |  |  |
| $800 D$ | EF |  |  |  |  |

